

ISO7741Tx-Q1 Automotive, High-Speed, Reinforced Quad-Channel Digital Isolator With Integrated Transformer Driver

1 Features

- Qualified for automotive applications AEC-Q100 qualified with the following results:
- Device temperature Grade 1: –40°C to 125°C ambient operating temperature
- [Functional Safety-Capable](#)
 - Documentation available to aid ISO 26262 system design
- 100Mbps data rate
- Robust isolation barrier:
 - >30-year projected lifetime at 1500V_{RMS} working voltage
 - 5000V_{RMS} rating
 - Up to 12.8kV surge capability
 - ±150kV/μs typical CMTI for digital isolator
- Wide supply range: 2.25V to 5.5V
- 2.25V to 5.5V level translation
- Default output *high* (ISO7741T) and *low* (ISO7741FT) options
- Low power consumption, typical 1.5mA per channel at 1Mbps
- Low propagation delay: 10.7ns typical (5V Supplies)
- Robust electromagnetic compatibility (EMC)
 - System-level ESD, EFT, and surge immunity
- Push-pull driver for transformers
- High output drive: 0.7A at 5V supply
- Low R_{ON} 0.4Ω max at 4.5V supply
- Spread spectrum clocking
- 1.7A Current-limit
- Safety-related certifications (Pending):
 - DIN EN IEC 60747-17 (VDE 0884-17)
 - UL 1577 component recognition program
 - IEC 61010-1, IEC 62368-1, IEC 60601-1, and GB 4943.1 certifications

2 Applications

Hybrid, electric and powertrain system (EV/HEV)

- [Battery management system \(BMS\)](#)
- [On-board charger](#)
- [Traction inverter and motor control](#)
- [DC/DC converter](#)

Body electronics

- [Automotive parking heater module](#)
- [HVAC control module](#)

3 Description

The ISO7741T-Q1 is a grade 1, high-performance, quad-channel digital isolator with integrated transformer driver with 5000V_{RMS} isolation ratings per UL 1577. This reinforced insulation ratings according to VDE, CSA, TUV and CQC.

The ISO7741T-Q1 device provide high electromagnetic immunity and low emissions at low power consumption, while isolating CMOS or LVC MOS digital I/Os. Each isolation channel has a logic input and output buffer separated by a double capacitive silicon dioxide (SiO₂) insulation barrier. If the input power or signal is lost, default output is *high* for devices without suffix F and *low* for devices with suffix F. See the [Device Functional Modes](#) section for further details.

The ISO7741T-Q1 has a low-noise, low-EMI push-pull transformer driver, specifically designed for small form factor, isolated power supplies. The device drives low-profile, center-tapped transformers from a 2.25V to 5.5V DC power supply. Very low noise and EMI are achieved by slew rate control of the output switch voltage and through Spread Spectrum Clocking (SSC). The ISO7741T-Q1 consists of an oscillator followed by a gate drive circuit that provides the complementary output signals to drive ground-referenced N-channel power switches. The device includes two 1A Power-MOSFET switches to provide start-up under heavy loads. The internal protection features include a 1.7A current limiting, under-voltage lockout, thermal shutdown, and break-before-make circuitry. ISO7741T-Q1 include a soft-start feature that prevents high inrush current during power up with large load capacitors. ISO7741TA-Q1 has a 160kHz internal oscillator for applications that need to minimize emissions whereas ISO7741TB-Q1 has a 420kHz internal oscillators for applications that require higher efficiency and smaller transformer size.

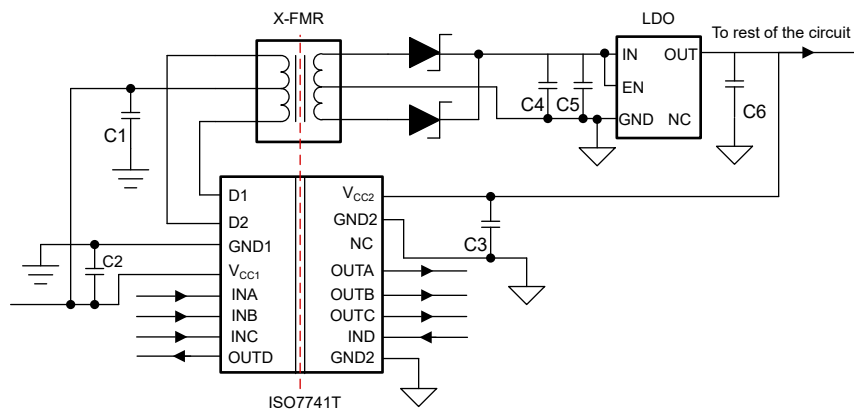
The ISO7741Tx-Q1 is available in a 16-pin DW package. The device operation is characterized for a temperature range from –40°C to 125°C.



Package Information

PART NUMBER	PACKAGE ⁽¹⁾	BODY SIZE (NOM)	PACKAGE SIZE ⁽²⁾
ISO7741Tx-Q1	DW (16)	10.30mm × 7.50mm	10.30mm × 10.30mm

- (1) For all available packages, see [Section 11](#).
(2) The package size (length × width) is a nominal value and includes pins, where applicable.



Simplified Schematic

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4 Pin Configuration and Functions

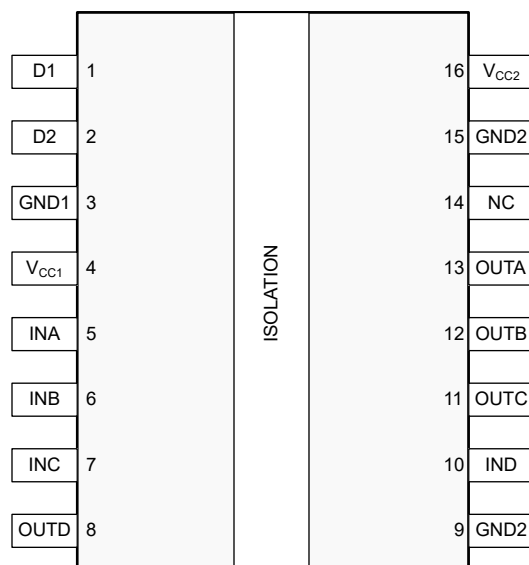


Figure 4-1. ISO7741T DW Package Top View

Pin Functions

NAME	PIN	Type ⁽¹⁾	DESCRIPTION
D1	1	O	Open drain output of the first power MOSFETs. Typically connected to the outer terminals of the center tap transformer. Because large currents flow through these pins, the external traces must be kept short.
D2	2	O	Open drain output of the second power MOSFETs. Typically connected to the outer terminals of the center tap transformer. Because large currents flow through these pins, the external traces must be kept short.
GND1	3	—	Ground connection for V _{CC1}
GND2	9	—	Ground connection for V _{CC2}
	15		
INA	5	I	Input, channel A
INB	6	I	Input, channel B
INC	7	I	Input, channel C
IND	10	I	Input, channel D
NC	14	—	Not connected
OUTA	13	O	Output, channel A
OUTB	12	O	Output, channel B
OUTC	11	O	Output, channel C
OUTD	8	O	Output, channel D
V _{CC1}	4	—	Power supply, side 1
V _{CC2}	16	—	Power supply, side 2

(1) I = Input, O = Output

5 Specifications

5.1 Absolute Maximum Ratings

See⁽¹⁾

		MIN	MAX	UNIT
V_{CC1}, V_{CC2}	Supply voltage ⁽²⁾	-0.5	6	V
V	Voltage at INx, OUTx	-0.5	$V_{CCX} + 0.5$ ⁽³⁾	V
D1, D2	Output switch voltage		16	V
$I_{(D1)PK}, I_{(D2)PK}$	Peak output switch current		2.4	A
I_O	Output current	-15	15	mA
T_J	Junction temperature	-40	150	°C
T_{stg}	Storage temperature	-65	150	°C

- (1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If used outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.
- (2) All voltage values except differential I/O bus voltages are with respect to the local ground terminal (GND1 or GND2) and are peak voltage values.
- (3) Maximum voltage must not exceed 6V.

5.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 HBM ESD Classification Level 3A ⁽¹⁾	±6000	V
		Charged-device model (CDM), per AEC Q100-011 V CDM ESD Classification Level C6	±1500	V

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

5.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V_{CC1}, V_{CC2}	Supply Voltage		2.25		5.5	V
$V_{CC(UVLO+)}$	UVLO threshold when supply voltage is rising			2	2.25	V
$V_{CC(UVLO-)}$	UVLO threshold when supply voltage is falling		1.7	1.8		V
$V_{HYS(UVLO)}$	Supply voltage UVLO hysteresis		100	200		mV
I_{D1}, I_{D2}	Output switch current - Primary side	$2.25V \leq V_{CC} \leq 2.8V$			0.7	A
I_{D1}, I_{D2}	Output switch current - Primary side	$2.8V < V_{CC} \leq 5.5V$			0.7	A
I_{OH}	High level output current	$V_{CCO} = 5V$ ⁽¹⁾	-4			mA
		$V_{CCO} = 3.3V$	-2			
		$V_{CCO} = 2.5V$	-1			
I_{OL}	Low level output current	$V_{CCO} = 5V$			4	mA
		$V_{CCO} = 3.3V$			2	
		$V_{CCO} = 2.5V$			1	
V_{IH}	High level Input voltage		$0.7 \times V_{CCI}$ ⁽¹⁾		V_{CCI}	V
V_{IL}	Low level Input voltage		0	$0.3 \times V_{CCI}$		V
DR	Data Rate ⁽²⁾		0		100	Mbps
T_A	Ambient temperature		-40	25	125	°C

(1) V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} .

(2) 100Mbps is the maximum specified data rate, although higher data rates are possible.

5.4 Thermal Information

THERMAL METRIC ⁽¹⁾		ISO7741T-Q1	UNIT
		DW (SOIC)	
		16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	64.0	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	25.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	33.3	°C/W
Ψ_{JT}	Junction-to-top characterization parameter	8.4	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	32.7	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	NA	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application note.

5.5 Power Ratings

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ISO7741T-Q1						
P_D	Maximum power dissipation (both sides)	$V_{CC1} = V_{CC2} = 5.5V$, $T_J = 150^\circ C$, $C_L = 15pF$, Input a 50MHz 50% duty cycle square wave, $I_{D1}, I_{D2} = 700mA$			401	mW
P_{D1}	Maximum power dissipation (side-1)	$V_{CC1} = 5.5V$, $T_J = 150^\circ C$, $C_L = 15pF$, Input a 50MHz 50% duty cycle square wave, $I_{D1}, I_{D2} = 700mA$			266	mW
P_{D2}	Maximum power dissipation (side-2)	$V_{CC2} = 5.5V$, $T_J = 150^\circ C$, $C_L = 15pF$, Input a 50MHz 50% duty cycle square wave, $I_{D1}, I_{D2} = 700mA$			135	mW

5.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
			DW-16	
CLR	External clearance ⁽¹⁾	Shortest terminal-to-terminal distance through air	>8	mm
CPG	External creepage ⁽¹⁾	Shortest terminal-to-terminal distance across the package surface	>8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	>17	μm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	>600	V
	Material group	According to IEC 60664-1	I	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 300V _{RMS}	I-IV	
		Rated mains voltage ≤ 600V _{RMS}	I-IV	
		Rated mains voltage ≤ 1000V _{RMS}	I-III	
DIN EN IEC 60747-17 (VDE 0884-17) ⁽²⁾				
V _{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	2121	V _{PK}
V _{IOWM}	Maximum working isolation voltage	AC voltage; Time dependent dielectric breakdown (TDDb) Test	1500	V _{RMS}
		DC voltage	2121	V _{DC}
V _{IOTM}	Maximum transient isolation voltage	V _{TEST} = V _{IOTM} , t = 60s (qualification); V _{TEST} = 1.2 x V _{IOTM} , t= 1s (100% production)	7071	V _{PK}
V _{IMP}	Maximum impulse voltage ⁽³⁾	Tested in air, 1.2/50-μs waveform per IEC 62368-1	9846	V _{PK}
V _{IOSM}	Maximum surge isolation voltage ⁽⁴⁾	V _{IOSM} ≥ 1.3 x V _{IMP} ; Tested in oil (qualification test), 1.2/50μs waveform per IEC 62368-1	12800	V _{PK}
q _{pd}	Apparent charge ⁽⁵⁾	Method a, After Input output safety test subgroup 2/3, V _{ini} = V _{IOTM} , t _{ini} = 60s; V _{pd(m)} = 1.2 x V _{IORM} , t _m = 10s	≤5	pC
		Method a, After environmental tests subgroup 1, V _{ini} = V _{IOTM} , t _{ini} = 60s; V _{pd(m)} = 1.6 x V _{IORM} , t _m = 10s	≤5	
		Method b: At routine test (100% production); V _{ini} = 1.2 x V _{IOTM} , t _{ini} = 1s; V _{pd(m)} = 1.875 x V _{IORM} , t _m = 1s (method b1) or V _{pd(m)} = V _{ini} , t _m = t _{ini} (method b2)	≤5	
C _{IO}	Barrier capacitance, input to output ⁽⁶⁾	V _{IO} = 0.4 x sin (2πft), f = 1MHz	≅1	pF
R _{IO}	Isolation resistance ⁽⁶⁾	V _{IO} = 500V, T _A = 25°C	>10 ¹²	Ω
		V _{IO} = 500V, 100°C ≤ T _A ≤ 125°C	>10 ¹¹	
		V _{IO} = 500V at T _S = 150°C	>10 ⁹	
	Pollution degree		2	
	Climatic category		55/125/21	
UL 1577				
V _{ISO}	Maximum withstanding isolation voltage	V _{TEST} = V _{ISO} , t = 60s (qualification), V _{TEST} = 1.2 x V _{ISO} , t = 1s (100% production)	5000	V _{RMS}

- (1) Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed-circuit board are used to help increase these specifications.
- (2) This coupler is suitable for *safe electrical insulation* only within the safety ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- (3) Testing is carried out in air to determine the surge immunity of the package.
- (4) Testing is carried out in oil to determine the intrinsic surge immunity of the isolation barrier.
- (5) Apparent charge is electrical discharge caused by a partial discharge (pd).

- (6) All pins on each side of the barrier tied together creating a two-terminal device.

5.7 Safety-Related Certifications

VDE	CSA	UL	CQC	TUV
Plan to certify according to DIN EN IEC 60747-17 (VDE 0884-17)	Plan to certify according to IEC 62368-1 and IEC 60601-1	Plan to certify according to UL 1577 Component Recognition Program	Plan to certify according to GB 4943.1	Plan to certify according to EN 61010-1 and EN 62368-1
Maximum transient isolation voltage, 7071 V _{PK} (DW-16, Reinforced); Maximum repetitive peak isolation voltage, 2121 V _{PK} (DW-16, Reinforced); Maximum surge isolation voltage, 12800 V _{PK} (DW-16, Reinforced)	Reinforced insulation per CSA 62368-1 and IEC 62368-1, 600 V _{RMS} (DW-16) max working voltage (pollution degree 2, material group I); 2 MOPP (Means of Patient Protection) per CSA 60601-1 and IEC 60601-1, 250 V _{RMS} (DW-16) max working voltage	DW-16: Single protection, 5000 V _{RMS} ;	DW-16: Reinforced Insulation, Altitude ≤ 5000 m, Tropical Climate, 700 V _{RMS} maximum working voltage;	5000 V _{RMS} (DW-16) Reinforced insulation per EN 61010-1 up to working voltage of 600 V _{RMS} (DW-16); 5000 V _{RMS} (DW-16) Reinforced insulation per EN 62368-1 up to working voltage of 600 V _{RMS} (DW-16)
Certificate planned	Certificate planned	Certificate planned	Certificate planned	Certificate planned

5.8 Safety Limiting Values

Safety limiting⁽¹⁾ intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry. A failure of the I/O can allow low resistance to ground or the supply and, without current limiting, dissipate sufficient power to overheat the die and damage the isolation barrier potentially leading to secondary system failures.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
DW-16 PACKAGE						
I _S	Safety input, output, or supply current	R _{θJA} = 64°C/W, V _I = 5.5V, T _J = 150°C, T _A = 25°C, I _L = 700mA see Figure 5-1			319	mA
		R _{θJA} = 64°C/W, V _I = 3.6V, T _J = 150°C, T _A = 25°C, I _L = 700mA see Figure 5-1			480	
		R _{θJA} = 64°C/W, V _I = 2.75V, T _J = 150°C, T _A = 25°C, I _L = 700mA see Figure 5-1			630	
P _S	Safety input, output, or total power	R _{θJA} = 64°C/W, T _J = 150°C, T _A = 25°C see Figure 5-2			1953	mW
T _S	Maximum safety temperature				150	°C

- (1) The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S must not be exceeded. These limits vary with the ambient temperature, T_A.
The junction-to-air thermal resistance, R_{θJA}, in [Section 5.4](#) is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:
T_J = T_A + R_{θJA} × P, where P is the power dissipated in the device.
T_{J(max)} = T_S = T_A + R_{θJA} × P_S, where T_{J(max)} is the maximum allowed junction temperature.
P_S = I_S × V_I, where V_I is the maximum input voltage.

5.9 Electrical Characteristics Transformer

over full-range of recommended operating conditions, unless otherwise noted. All typical values are at $T_A = 25^\circ\text{C}$, $V_{CC} = 5\text{V}$.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
CLK						
F_{SW}	D1, D2 average switching Frequency (ISO7741TA-Q1)	$R_L = 50\Omega$ to V_{CC1}	138	160	203	kHz
	D1, D2 average switching Frequency (ISO7741TB-Q1)	$R_L = 50\Omega$ to V_{CC1}	363	424	517	kHz
OUTPUT STAGE						
DMM	Average ON time mismatch between D1 and D2	$R_L = 50\Omega$	0%			
$R_{(ON)}$	Output switch on resistance	$V_{CC} = 4.5\text{V}$, I_{D1} , $I_{D2} = 700\text{mA}$	0.22	0.4	Ω	
		$V_{CC} = 2.8\text{V}$, I_{D1} , $I_{D2} = 700\text{mA}$	0.24	0.45	Ω	
		$V_{CC} = 2.25\text{V}$, I_{D1} , $I_{D2} = 0.5\text{A}$	0.26	0.5	Ω	
$V_{(SLEW)}$	Voltage slew rates on D1 and D2 for ISO7741TA-Q1	$R_L = 50\Omega$ to V_{CC1}	48		$\text{V}/\mu\text{s}$	
$I_{(SLEW)}$	Current slew rates at D1 and D2 for ISO7741TA-Q1	$R_L = 5\Omega$ through transformer	11		$\text{A}/\mu\text{s}$	
$V_{(SLEWHF)}$	Voltage slew rates on D1 and D2 for ISO7741TB-Q1	$R_L = 50\Omega$ to V_{CC1}	152		$\text{V}/\mu\text{s}$	
$I_{(SLEWHF)}$	Current slew rates at D1 and D2 for ISO7741TB-Q1	$R_L = 5\Omega$ through transformer	41		$\text{A}/\mu\text{s}$	
I_{LIM}	Current clamp limit ($2.8\text{V} < V_{CC} \leq 5.5\text{V}$)		1.42	1.75	2.15	A
	Current clamp limit ($2.25\text{V} \leq V_{CC} \leq 2.8\text{V}$)		0.65		1.85	A
THERMAL SHUT DOWN						
T_{SD+}	T_{SD} turn on temperature		154	168	181	$^\circ\text{C}$
T_{SD-}	T_{SD} turn off temperature		135	150	166	$^\circ\text{C}$
T_{SD-}	T_{SD} hysteresis		13	17	$^\circ\text{C}$	

5.10 Electrical Characteristics—5V Supply

$V_{CC1} = V_{CC2} = 5V \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	$I_{OH} = -4mA$	$V_{CCO} - 0.4$ ⁽¹⁾	4.8		V
V_{OL}	Low-level output voltage	$I_{OL} = 4mA$		0.2	0.4	V
$V_{IT+(IN)}$	Rising input switching threshold			$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$	V
$V_{IT-(IN)}$	Falling input switching threshold		$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis		$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
I_{IH}	High-level input current	$V_{IH} = V_{CCI}$ ⁽¹⁾ at INx			10	μA
I_{IL}	Low-level input current	$V_{IL} = 0V$ at INx	-10			μA
CMTI ⁽³⁾	Common mode transient immunity	$V_I = V_{CCI}$ or 0V, $V_{CM} = 1200V$	100	150		kV/ μs
C_I	Input Capacitance ⁽²⁾	$V_I = V_{CC}/2 + 0.4 \times \sin(2\pi ft)$, $f = 1MHz$, $V_{CC} = 5V$		2		pF

- (1) $V_{CCI} =$ Input-side V_{CC} ; $V_{CCO} =$ Output-side V_{CC}
(2) Measured from input pin to same side ground.
(3) CMTI is measured only on digital isolator independent of transformer driver

5.11 Supply Current Characteristics—5V Supply

$V_{CC1} = V_{CC2} = 5V \pm 10\%$, $R_L = 50\Omega$ to V_{CC1} , (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT			
ISO7741TA-Q1										
Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}		3	4.2	mA			
			I_{CC2}		2.4	3.7				
	$V_I = 0V_I$ (ISO7741T) $V_I = V_{CCI}$ (ISO7741FT)		I_{CC1}		6.3	7.8				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		4.7	6.1				
			I_{CC2}		3.2	4.6				
		10Mbps	I_{CC1}		5.4	6.8				
			I_{CC2}		4.8	6.5				
		100Mbps	I_{CC1}		11.5	13.7				
			I_{CC2}		20	23.5				
		ISO7741TB-Q1								
		Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}			3.6	5	mA
I_{CC2}					2.4	3.7				
$V_I = 0V_I$ (ISO7741T) $V_I = V_{CCI}$ (ISO7741FT)			I_{CC1}		6.9	8.6				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		5.3	6.9				
			I_{CC2}		3.2	4.6				
		10Mbps	I_{CC1}		5.9	7.7				
			I_{CC2}		4.8	6.5				
		100Mbps	I_{CC1}		12.1	14.6				
			I_{CC2}		20	23.5				

(1) V_{CCI} = Input-side V_{CC}

5.12 Electrical Characteristics—3.3V Supply

$V_{CC1} = V_{CC2} = 3.3V \pm 10\%$, $R_L = 50\Omega$ to V_{CC1} , (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage $I_{OH} = -2mA$	$V_{CCO} - 0.3$ ⁽¹⁾	3.2		V
V_{OL}	Low-level output voltage $I_{OL} = 2mA$		0.1	0.3	V
$V_{IT+(IN)}$	Rising input switching threshold		$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$ ⁽¹⁾	V
$V_{IT-(IN)}$	Falling input switching threshold	$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis	$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
I_{IH}	High-level input current $V_{IH} = V_{CCI}$ ⁽¹⁾ at INx			10	μA
I_{IL}	Low-level input current $V_{IL} = 0V$ at INx	-10			μA
CMTI ⁽²⁾	Common mode transient immunity $V_I = V_{CCI}$ or 0V, $V_{CM} = 1200V$	100	150		kV/ μs

(1) V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC}

(2) CMTI is measured only on digital isolator independent of transformer driver

5.13 Supply Current Characteristics—3.3V Supply

$V_{CC1} = V_{CC2} = 3.3V \pm 10\%$, $R_L = 50\Omega$ to V_{CC1} , (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT			
ISO7741TA-Q1										
Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}		2.8	3.9	mA			
			I_{CC2}		2.3	3.7				
	$V_I = 0V_I$ (ISO7741T) ⁽¹⁾ $V_I = V_{CCI}$ (ISO7741FT)		I_{CC1}		6.1	7.5				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		4.5	5.7				
			I_{CC2}		3.1	4.5				
		10Mbps	I_{CC1}		4.9	6.3				
			I_{CC2}		4.2	5.9				
		100Mbps	I_{CC1}		9.2	11.1				
			I_{CC2}		15.1	18.2				
		ISO7741TB-Q1								
		Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}			3.1	4.4	mA
I_{CC2}					2.3	3.7				
$V_I = 0V_I$ (ISO7741T) ⁽¹⁾ $V_I = V_{CCI}$ (ISO7741FT)			I_{CC1}		6.5	8				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		4.9	6.3				
			I_{CC2}		3.1	4.5				
		10Mbps	I_{CC1}		5.3	6.9				
			I_{CC2}		4.2	5.9				
		100Mbps	I_{CC1}		9.6	11.7				
			I_{CC2}		15.1	18.2				

(1) V_{CCI} = Input-side V_{CC}

5.14 Electrical Characteristics—2.5V Supply

$V_{CC1} = V_{CC2} = 2.5V \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	High-level output voltage	$I_{OH} = -1mA$	$V_{CCO} - 0.2^{(1)}$	2.45		V
V_{OL}	Low-level output voltage	$I_{OL} = 1mA$		0.05	0.2	V
$V_{IT+(IN)}$	Rising input switching threshold			$0.6 \times V_{CCI}$	$0.7 \times V_{CCI}$	V
$V_{IT-(IN)}$	Falling input switching threshold		$0.3 \times V_{CCI}$	$0.4 \times V_{CCI}$		V
$V_{I(HYS)}$	Input threshold voltage hysteresis		$0.1 \times V_{CCI}$	$0.2 \times V_{CCI}$		V
I_{IH}	High-level input current	$V_{IH} = V_{CCI}^{(1)}$ at INx			10	μA
I_{IL}	Low-level input current	$V_{IL} = 0V$ at INx	-10			μA
CMTI ⁽²⁾	Common mode transient immunity	$V_I = V_{CCI}$ or 0V, $V_{CM} = 1200V$	100	150		kV/ μs

(1) V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC}

(2) CMTI is measured only on digital isolator independent of transformer driver

5.15 Supply Current Characteristics—2.5V Supply

$V_{CC1} = V_{CC2} = 2.5V \pm 10\%$, $R_L = 50 \Omega$ to V_{CC1} , (over recommended operating conditions unless otherwise noted)

PARAMETER	TEST CONDITIONS		SUPPLY CURRENT	MIN	TYP	MAX	UNIT			
ISO7741TA-Q1										
Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}		2.7	3.7	mA			
			I_{CC2}		2.3	3.7				
	$V_I = 0V_I$ (ISO7741T) ⁽¹⁾ $V_I = V_{CCI}$ (ISO7741FT)		I_{CC1}		6	7.3				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		4.4	5.6				
			I_{CC2}		3	4.5				
		10Mbps	I_{CC1}		4.7	6.1				
			I_{CC2}		3.9	5.5				
		100Mbps	I_{CC1}		8	9.7				
			I_{CC2}		12.1	14.7				
		ISO7741TB-Q1								
		Supply current - DC signal	$V_I = V_{CCI}$ (ISO7741T) ⁽¹⁾ $V_I = 0V$ (ISO7741FT)		I_{CC1}			3	4.2	mA
I_{CC2}					2.3	3.7				
$V_I = 0V_I$ (ISO7741T) ⁽¹⁾ $V_I = V_{CCI}$ (ISO7741FT)			I_{CC1}		6.3	7.8				
			I_{CC2}		3.6	5.1				
Supply current - AC signal	All channels switching with square wave clock input; $C_L = 15pF$	1Mbps	I_{CC1}		4.7	6				
			I_{CC2}		3	4.5				
		10Mbps	I_{CC1}		5	6.5				
			I_{CC2}		3.9	5.5				
		100Mbps	I_{CC1}		8.3	10.1				
			I_{CC2}		12.1	14.7				

(1) V_{CCI} = Input-side V_{CC}

5.16 Switching Characteristics—5V Supply

$V_{CC1} = V_{CC2} = 5V \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time		6	10.7	17	ns
PWD	Pulse width distortion ⁽¹⁾ $ t_{PHL} - t_{PLH} $			0	6.2	ns
$t_{sk(o)}$	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			4	ns
$t_{sk(pp)}$	Part-to-part skew time ⁽³⁾				4.4	ns
t_r	Output signal rise time			2.4	3.9	ns
t_f	Output signal fall time			2.4	3.9	ns
t_{DO}	Default output delay time from input power loss	Measured from the time V_{CC} goes below 1.7V.		0.1	0.3	μs
t_{ie}	Time interval error	$2^{16} - 1$ PRBS data at 100Mbps		0.8		ns
t_{BBM}	Break-before-make time (ISO7741TA-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC}		115		ns
t_{BBM}	Break-before-make time (ISO7741TB-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC}		90		ns
t_{SS}	Soft-start time (SN6505A-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	2.2	8	ms
t_{SS}	Soft-start time (SN6505B-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	4.25	8	ms
$t_{SSdelay}$	Soft-start time delay	From power up to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	3.5	8.5	18	ms

(1) Also known as pulse skew.

(2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

(3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

5.17 Switching Characteristics—3.3V Supply

$V_{CC1} = V_{CC2} = 3.3V \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time		6	12	18.5	ns
PWD	Pulse width distortion ⁽¹⁾ $ t_{PHL} - t_{PLH} $			0.1	6.2	ns
$t_{sk(o)}$	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			4.4	ns
$t_{sk(pp)}$	Part-to-part skew time ⁽³⁾				5	ns
t_r	Output signal rise time			1.3	3	ns
t_f	Output signal fall time			1.3	3	ns
t_{DO}	Default output delay time from input power loss	Measured from the time V_{CC} goes below 1.7V.		0.1	0.3	μs
t_{ie}	Time interval error	$2^{16} - 1$ PRBS data at 100Mbps		0.9		ns
t_{BBM}	Break-before-make time (ISO7741TA-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC}		115		ns
t_{BBM}	Break-before-make time (ISO7741TB-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC}		90		ns
t_{SS}	Soft-start time (SN6505A-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	2.2	8	ms
t_{SS}	Soft-start time (SN6505B-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	4.25	8	ms
$t_{SSdelay}$	Soft-start time delay	From power up to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	3.5	8.5	18	ms

- (1) Also known as pulse skew.
 (2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.
 (3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

5.18 Switching Characteristics—2.5V Supply

 $V_{CC1} = V_{CC2} = 2.5V \pm 10\%$ (over recommended operating conditions unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
t_{PLH}, t_{PHL}	Propagation delay time		7.5	13	21	ns
PWD	Pulse width distortion ⁽¹⁾ $ t_{PHL} - t_{PLH} $			0.2	6.2	ns
$t_{sk(o)}$	Channel-to-channel output skew time ⁽²⁾	Same-direction channels			4.4	ns
$t_{sk(pp)}$	Part-to-part skew time ⁽³⁾				5.3	ns
t_r	Output signal rise time			1	3.5	ns
t_f	Output signal fall time			1	3.5	ns
t_{DO}	Default output delay time from input power loss	Measured from the time V_{CC} goes below 1.7V.		0.1	0.3	μs
t_{ie}	Time interval error	$2^{16} - 1$ PRBS data at 100Mbps		0.7		ns
t_{BBM}	Break-before-make time (ISO7741TA-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC} .		115		ns
t_{BBM}	Break-before-make time (ISO7741TB-Q1)	Measured as voltage with $R_L = 50\Omega$ to V_{CC} .		90		ns
t_{SS}	Soft-start time (SN6505A-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	2.2	8	ms
t_{SS}	Soft-start time (SN6505B-Q1)	10% to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	1	4.25	8	ms
$t_{SSdelay}$	Soft-start time delay	From power up to 90% transition time on V_{OUT} With transformer $C_{LOAD} = 40\mu F$ $R_L = 5\Omega$	3.5	8.5	18	ms

(1) Also known as pulse skew.

(2) $t_{sk(o)}$ is the skew between outputs of a single device with all driving inputs connected together and the outputs switching in the same direction while driving identical loads.

(3) $t_{sk(pp)}$ is the magnitude of the difference in propagation delay times between any terminals of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads.

5.19 Insulation Characteristics Curves

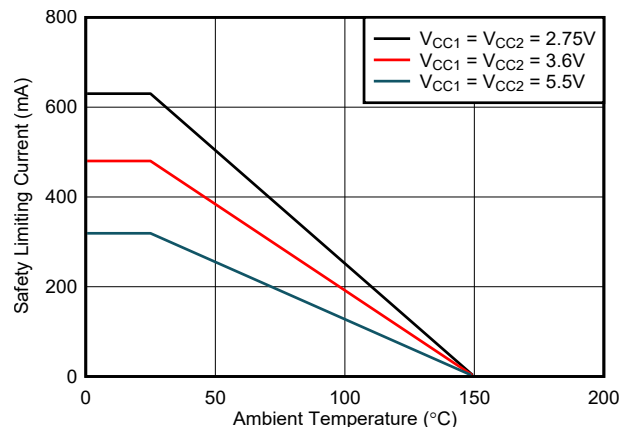


Figure 5-1. Thermal Derating Curve for Safety Limiting Current for DW-16 Package

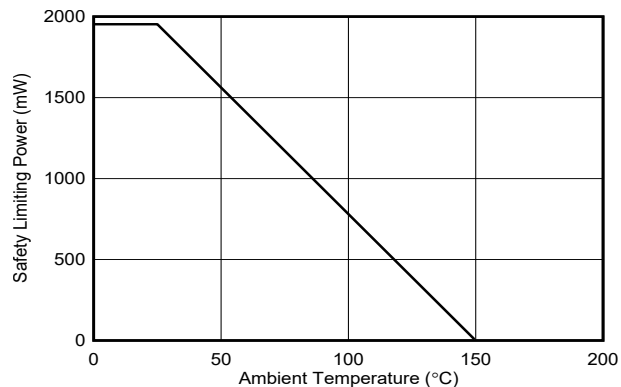


Figure 5-2. Thermal Derating Curve for Safety Limiting Power for DW-16 Package

5.20 Typical Characteristics

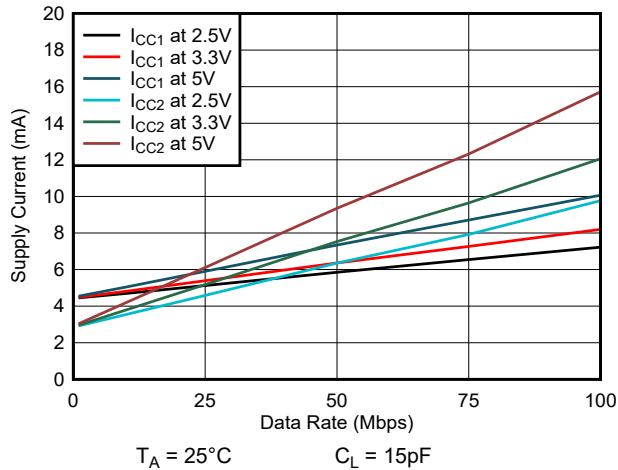


Figure 5-3. ISO7741Tx-Q1 Supply Current vs Data Rate (With 15pF Load)

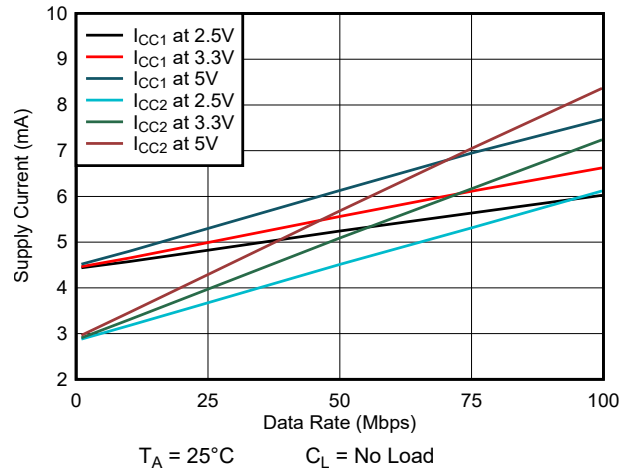


Figure 5-4. ISO7741Tx-Q1 Supply Current vs Data Rate (With No Load)

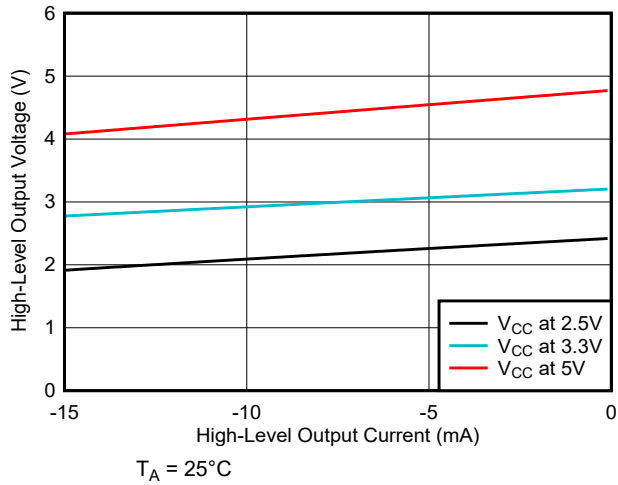


Figure 5-5. High-Level Output Voltage vs High-level Output Current

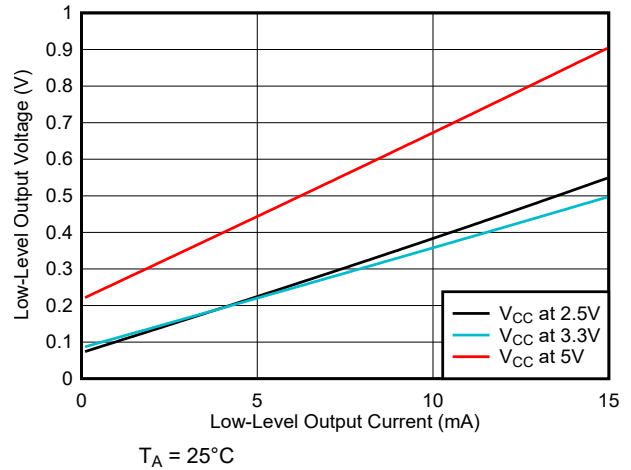


Figure 5-6. Low-Level Output Voltage vs Low-Level Output Current

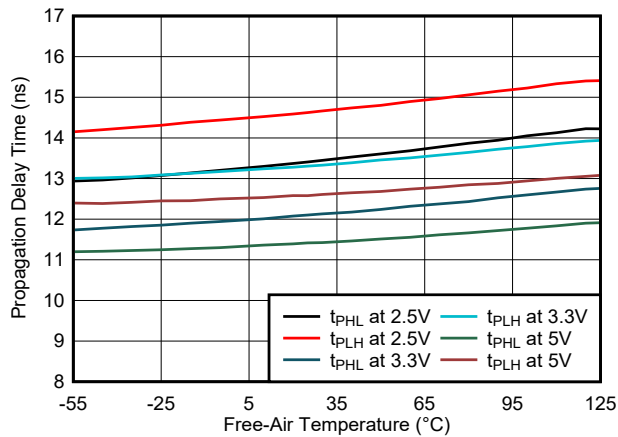


Figure 5-7. Propagation Delay Time vs Free-Air Temperature

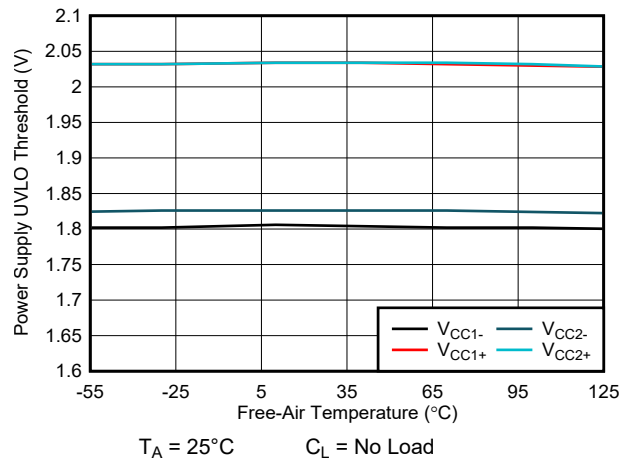
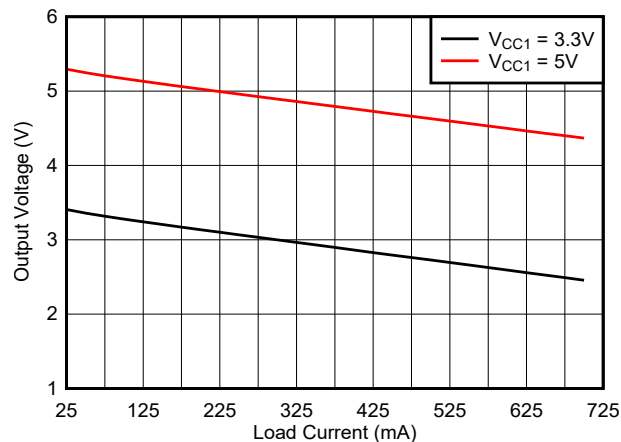
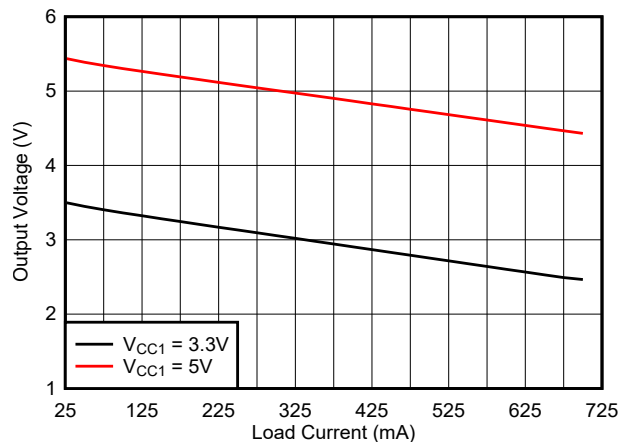


Figure 5-8. Power Supply Undervoltage Threshold vs Free-Air Temperature



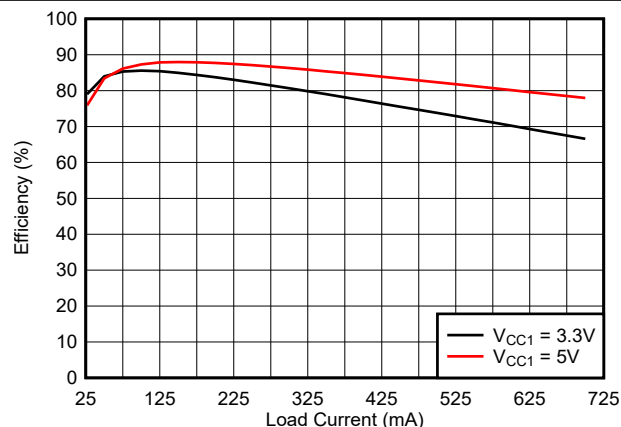
ISO7741TA-Q1 + Würth 750315240

Figure 5-9. ISO7741TA-Q1 Output Voltage vs Load Current



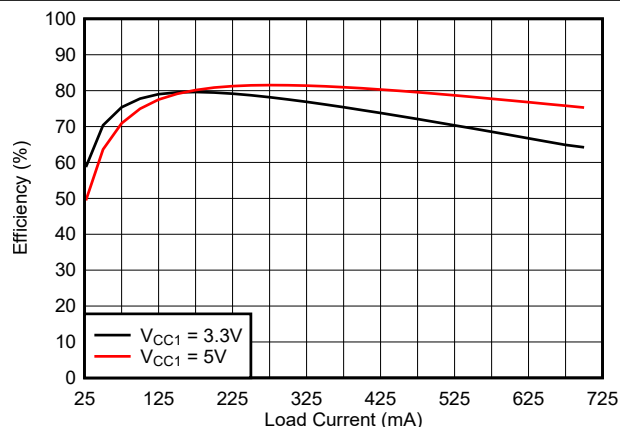
ISO7741TB-Q1 + Würth 750315371

Figure 5-10. ISO7741TB-Q1 Output Voltage vs Load Current



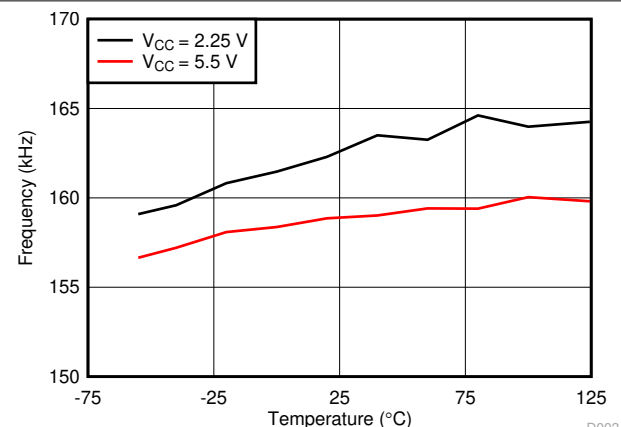
ISO7741TA-Q1 + Würth 750315240

Figure 5-11. ISO7741TA-Q1 Efficiency vs Load Current



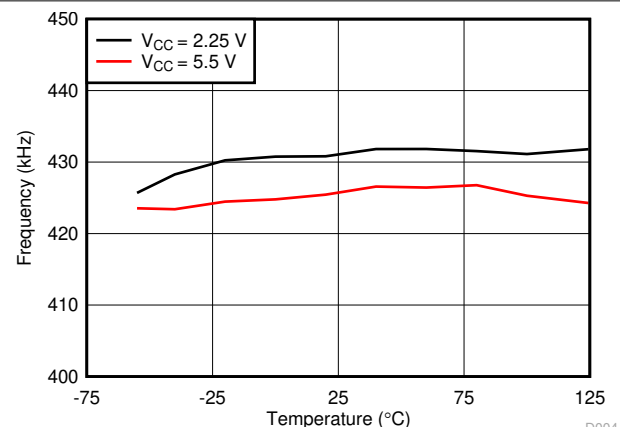
ISO7741TB-Q1 + Würth 750315371

Figure 5-12. ISO7741TB-Q1 Efficiency vs Load Current



D003

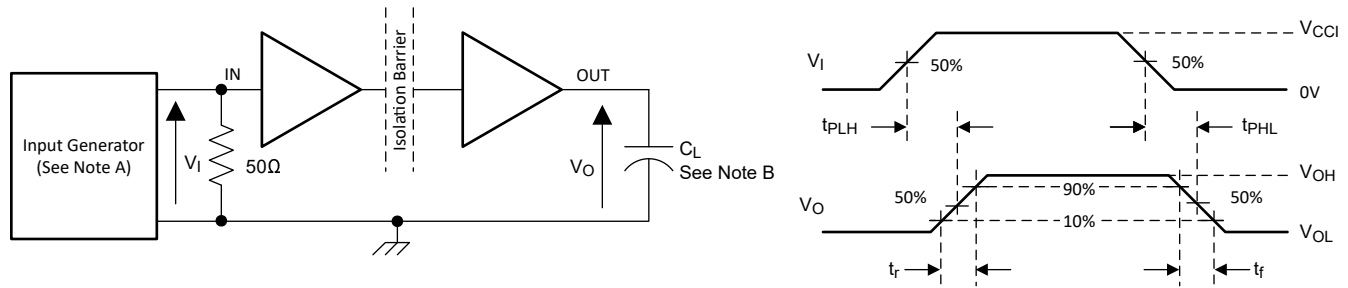
Figure 5-13. ISO7741TA-Q1 Frequency vs Free-Air Temperature



D004

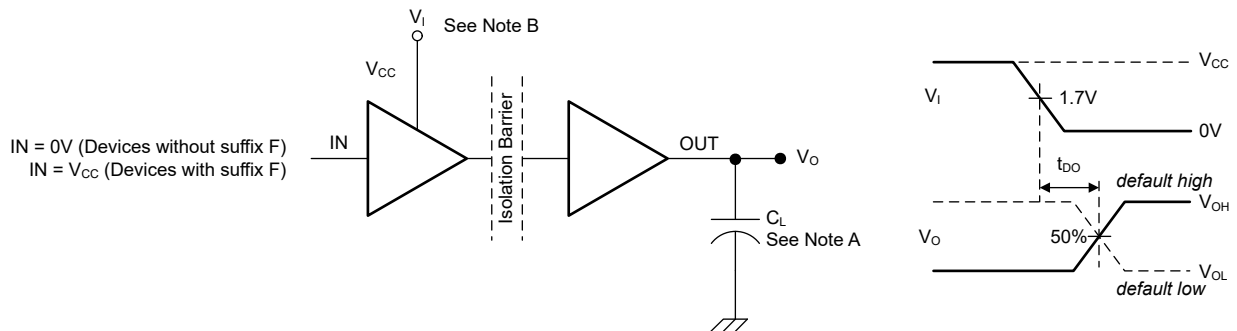
Figure 5-14. ISO7741TB-Q1 Frequency vs Free-Air Temperature

6 Parameter Measurement Information



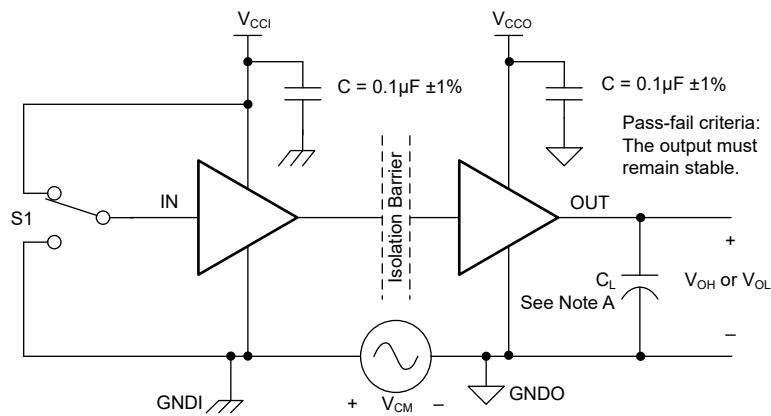
- A. The input pulse is supplied by a generator having the following characteristics: PRR $\leq$ 50kHz, 50% duty cycle, $t_r \leq$ 3ns, $t_f \leq$ 3ns, $Z_0 = 50\Omega$. At the input, 50 Ω resistor is required to terminate Input Generator signal. The 50 Ω resistor is not needed in actual application.
- B. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6-1. Switching Characteristics Test Circuit and Voltage Waveforms



- A. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.
- B. Power Supply Ramp Rate = 10mV/ns

Figure 6-2. Default Output Delay Time Test Circuit and Voltage Waveforms



- A. $C_L = 15$ pF and includes instrumentation and fixture capacitance within $\pm 20\%$.

Figure 6-3. Common-Mode Transient Immunity Test Circuit

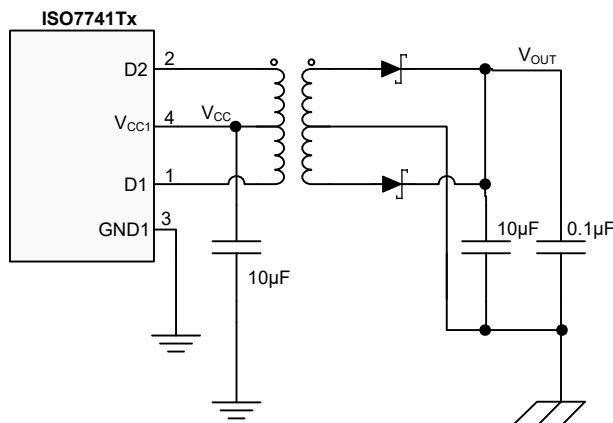


Figure 6-4. Measurement Circuit for Unregulated Output (TP1)

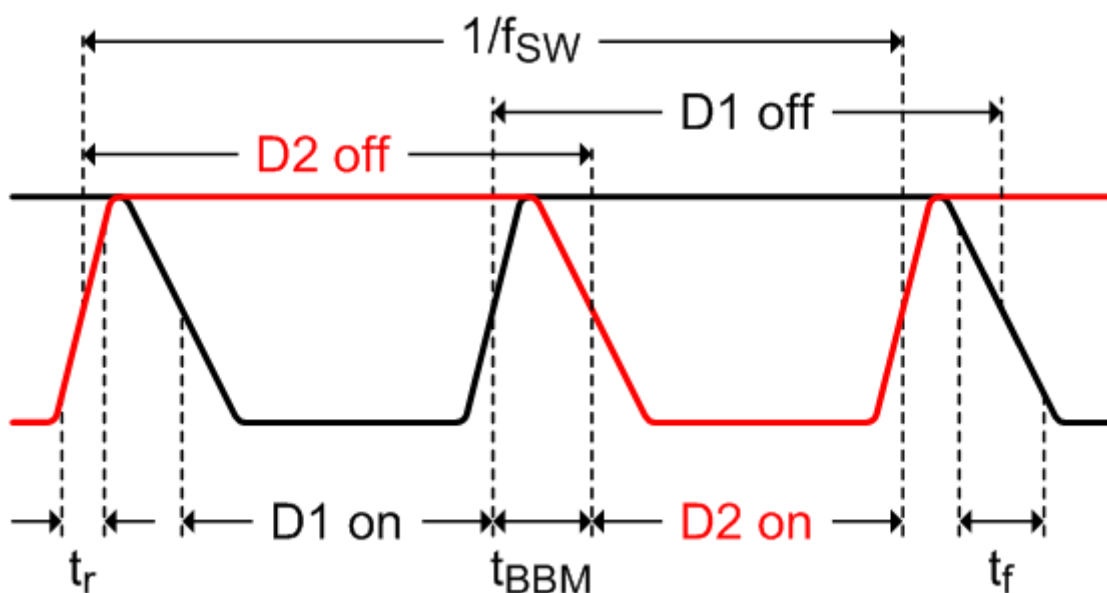


Figure 6-5. Timing Diagram

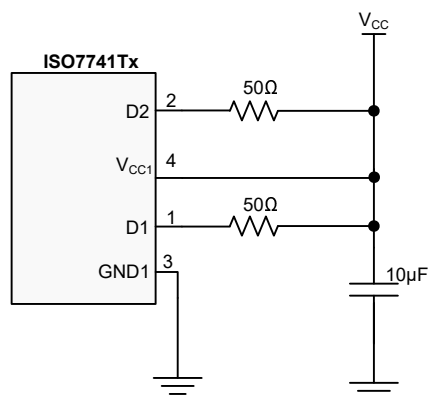


Figure 6-6. Test Circuit for FSW, V(slew), tBBM

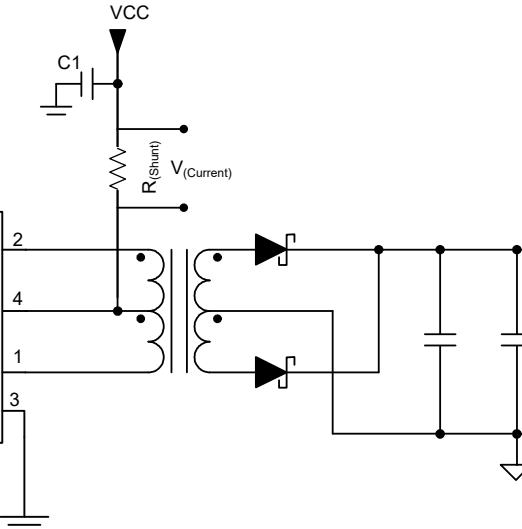


Figure 6-7. I(slew) Test Setup

7 Detailed Description

7.1 Overview

The ISO7741T devices have an ON-OFF keying (OOK) modulation scheme to transmit the digital data across a silicon dioxide based isolation barrier. The transmitter sends a high frequency carrier across the barrier to represent one digital state and sends no signal to represent the other digital state. The receiver demodulates the signal after advanced signal conditioning and produces the output through a buffer stage. The ISO7741T also incorporate advanced circuit techniques to maximize the CMTI performance and minimize the radiated emissions due to the high frequency carrier and IO buffer switching.

The transformer driver in ISO7741T is designed for low-cost, small form-factor, isolated DC/DC converters utilizing the push-pull topology. The device includes an oscillator that feeds a gate-drive circuit. The gate-drive, comprising a frequency divider and a break-before-make (BBM) logic, provides two complementary output signals which alternately turn the two output transistors on and off. The output frequency of the oscillator is divided down by two. A subsequent break-before-make logic inserts a dead-time between the high-pulses of the two signals. Before either one of the gates can assume logic high, the BBM logic provides a short time period during which both signals are low and both transistors are high-impedance. This short period, is required to avoid shorting out both ends of the primary. The resulting output signals, present the gate-drive signals for the output transistors.

The conceptual block diagram of a digital capacitive isolator with integrated transformer driver, [Figure 7-1](#) and [Figure 7-2](#), shows a functional block diagram of a typical channel and transformer driver.

7.2 Functional Block Diagram

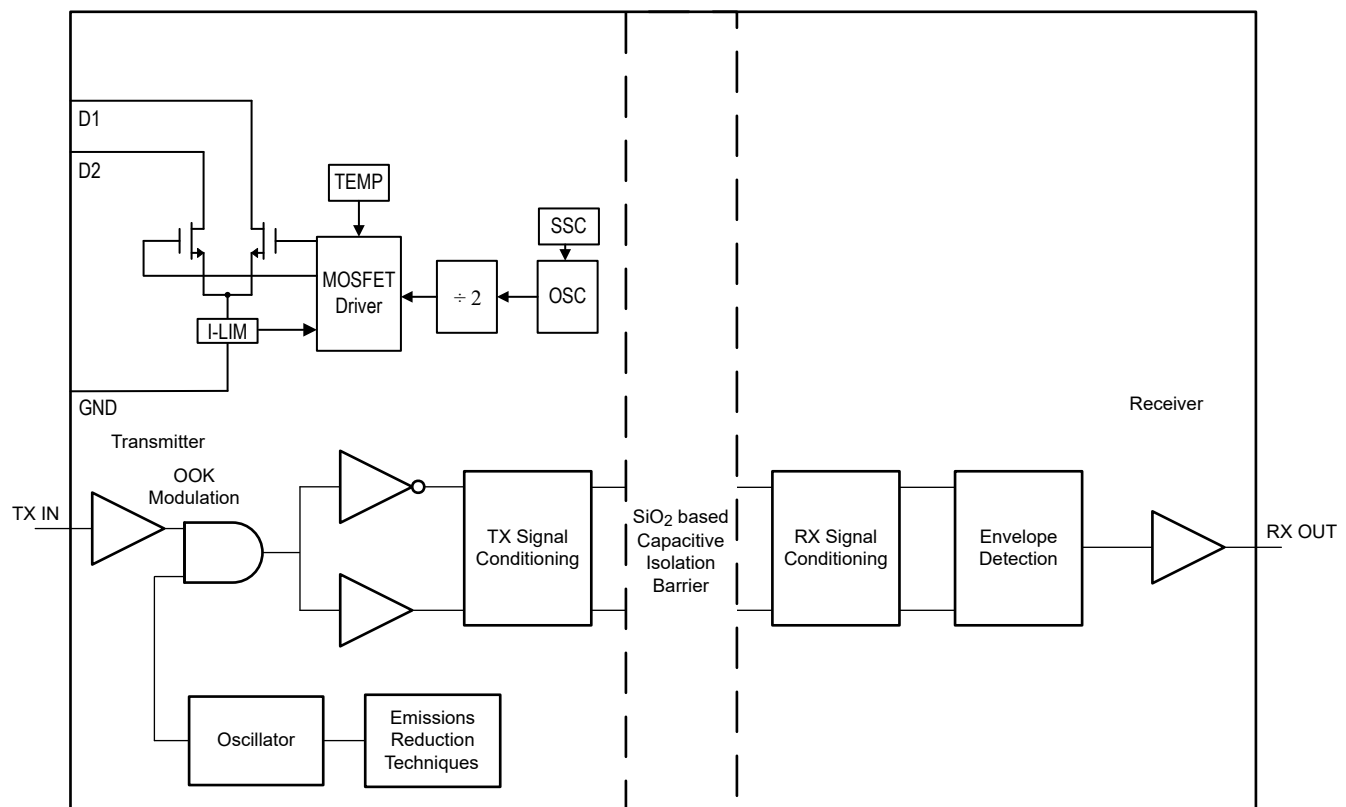


Figure 7-1. Conceptual Block Diagram of a Digital Capacitive Isolator

[Figure 7-2](#) shows a conceptual detail of how the ON-OFF keying scheme works.

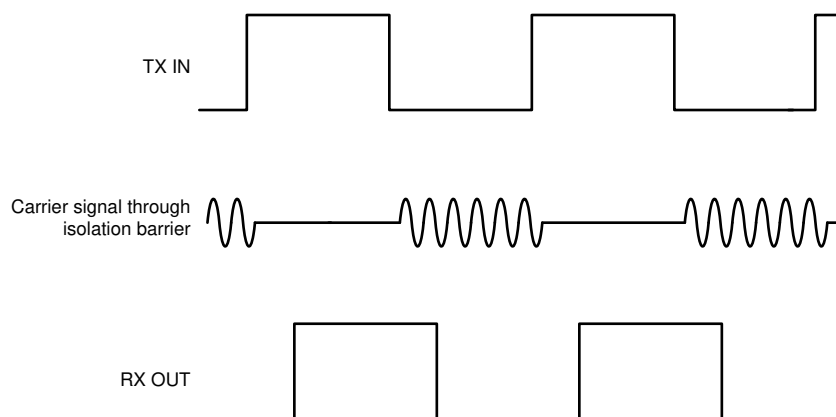


Figure 7-2. On-Off Keying (OOK) Based Modulation Scheme

7.3 Feature Description

Table 7-1. Device Features

PART NUMBER	TYPICAL SWITCHING FREQUENCY	MAXIMUM DATA RATE	DEFAULT OUTPUT	PACKAGE	RATED ISOLATION
ISO7741TA	160kHz	100Mbps	High	DW-16	5000V _{RMS} / 7071V _{PK}
ISO7741FTA	160kHz	100Mbps	Low	DW-16	5000V _{RMS} / 7071V _{PK}
ISO7741TB	424kHz	100Mbps	High	DW-16	5000V _{RMS} / 7071V _{PK}
ISO7741FTB	424kHz	100Mbps	Low	DW-16	5000V _{RMS} / 7071V _{PK}

See [Section 5.7](#) for detailed isolation ratings.

7.3.1 Electromagnetic Compatibility (EMC) Considerations

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 22. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the ISO7741T devices incorporate many chip-level design improvements for overall system robustness. Some of these improvements include:

- Robust ESD protection cells for input and output signal pins and inter-chip bond pads.
- Low-resistance connectivity of ESD cells to supply and ground pins.
- Enhanced performance of high voltage isolation capacitor for better tolerance of ESD, EFT and surge events.
- Bigger on-chip decoupling capacitors to bypass undesirable high energy signals through a low impedance path.
- PMOS and NMOS devices isolated from each other by using guard rings to avoid triggering of parasitic SCRs.
- Reduced common mode currents across the isolation barrier by providing purely differential internal operation.

7.3.2 Push-Pull Converter

Push-pull converters require transformers with center-taps to transfer power from the primary to the secondary (see [Figure 7-3](#)).

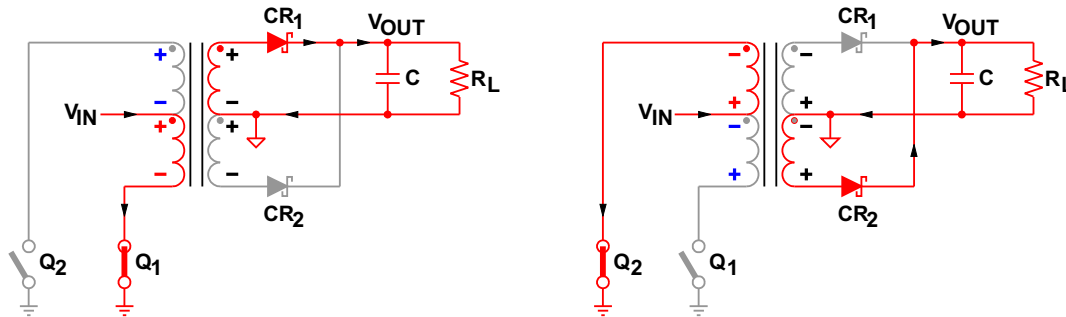


Figure 7-3. Switching Cycles of a Push-Pull Converter

When Q_1 conducts, V_{IN} drives a current through the lower half of the primary to ground, thus creating a negative voltage potential at the lower primary end with regards to the V_{IN} potential at the center-tap.

At the same time the voltage across the upper half of the primary is such that the upper primary end is positive with regards to the center-tap to maintain the previously established current flow through Q_2 , which now has turned high-impedance. The two voltage sources, each of which equaling V_{IN} , appear in series and cause a voltage potential at the open end of the primary of $2 \times V_{IN}$ with regards to ground.

Per dot convention the same voltage polarities that occur at the primary also occur at the secondary. The positive potential of the upper secondary end therefore forward biases diode CR_1 . The secondary current starting from the upper secondary end flows through CR_1 , charges capacitor C , and returns through the load impedance R_L back to the center-tap.

When Q_2 conducts, Q_1 goes high-impedance and the voltage polarities at the primary and secondary reverse. Now the lower end of the primary presents the open end with a $2 \times V_{IN}$ potential against ground. In this case CR_2 is forward biased while CR_1 is reverse biased and current flows from the lower secondary end through CR_2 , charging the capacitor and returning through the load to the center-tap.

7.3.3 Core Magnetization

Figure 7-4 shows the ideal magnetizing curve for a push-pull converter with B as the magnetic flux density and H as the magnetic field strength. When Q_1 conducts the magnetic flux is pushed from A to A' , and when Q_2 conducts the flux is pulled back from A' to A . The difference in flux and thus in flux density is proportional to the product of the primary voltage, V_P , and the time, t_{ON} , the voltage is applied to the primary: $B \cong V_P \times t_{ON}$.

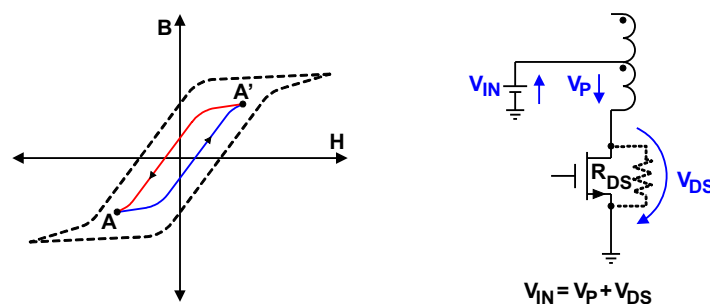


Figure 7-4. Core Magnetization and Self-Regulation Through Positive Temperature Coefficient of $R_{DS(on)}$

This volt-seconds (V - t) product is important as the product determines the core magnetization during each switching cycle. If the V - t products of both phases are not identical, an imbalance in flux density swing results with an offset from the origin of the B - H curve. If balance is not restored, the offset increases with each following cycle and the transformer slowly creeps toward the saturation region.

7.4 Device Functional Modes

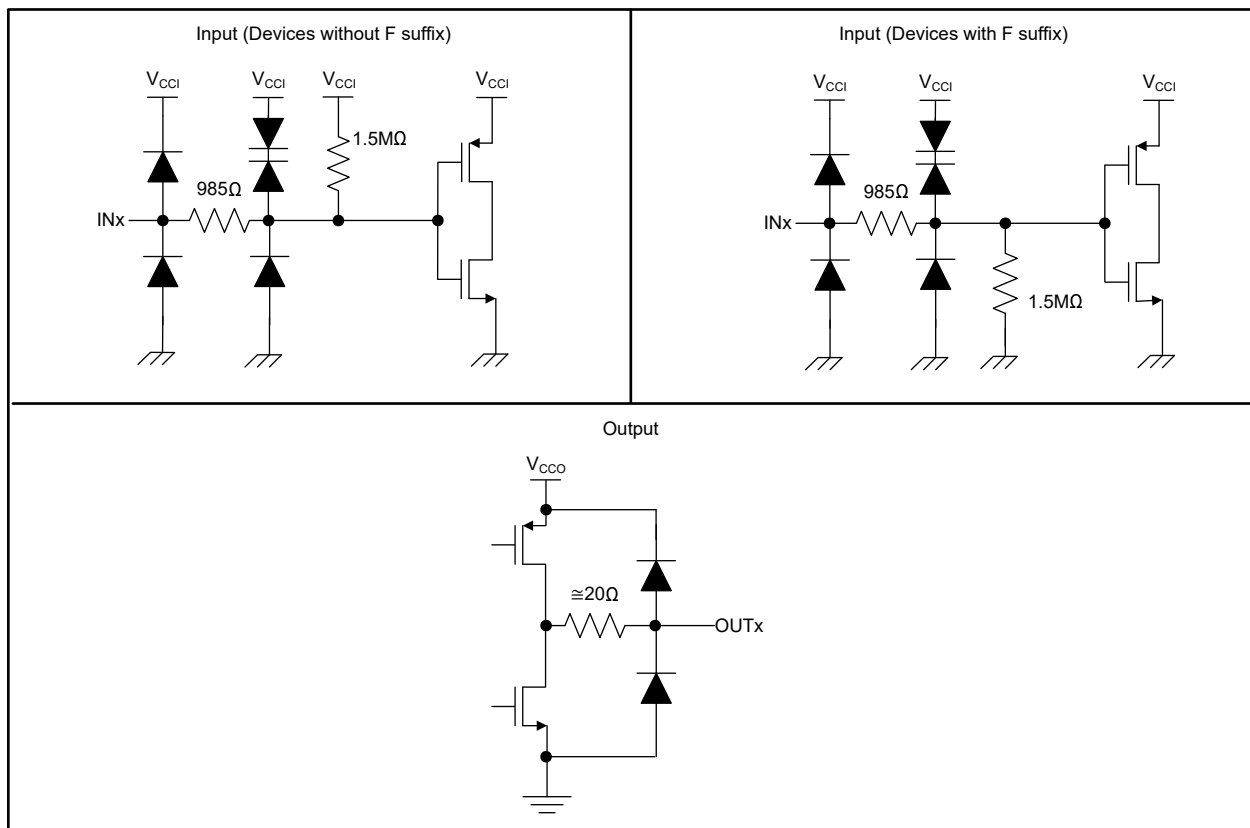
Table 7-2 lists the functional modes for the ISO7741T devices.

Table 7-2. Function Table

V_{CCI} ¹	V_{CCO}	INPUT (INx) ³	OUTPUT (OUTx)	COMMENTS
PU	PU	H	H	Normal Operation: A channel output assumes the logic state of the input.
		L	L	
		Open	Default	Default mode: When INx is open, the corresponding channel output goes to the default logic state. Default is <i>High</i> for ISO7741Tx and <i>Low</i> for ISO7741FTx.
PD	PU	X	Default	Default mode: When V_{CCI} is unpowered, a channel output assumes the logic state based on the selected default option. Default is <i>High</i> for ISO7741Tx and <i>Low</i> for ISO7741FTx. When V_{CCI} transitions from unpowered to powered-up, a channel output assumes the logic state of the input. When V_{CCI} transitions from powered-up to unpowered, channel output assumes the selected default state.
X	PD	X	Undetermined	When V_{CCO} is unpowered, a channel output is undetermined ² . When V_{CCO} transitions from unpowered to powered-up, a channel output assumes the logic state of the input.

1. V_{CCI} = Input-side V_{CC} ; V_{CCO} = Output-side V_{CC} ; PU = Powered up ($V_{CC} \geq 2.25V$); PD = Powered down ($V_{CC} \leq 1.7V$); X = Irrelevant; H = High level; L = Low level; Z = High Impedance
2. The outputs are in undetermined state when $1.7V < V_{CCI}$, $V_{CCO} < 2.25V$.
3. A strongly driven input signal can weakly power the floating V_{CC} through an internal protection diode and cause undetermined output.

7.4.1 Device I/O Schematics



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Figure 7-5. Device I/O Schematics

7.4.2 Start-Up Mode

When the supply voltage at V_{CC1} ramps up to 2.25V, the internal oscillator starts operating. The output stage begins switching but the amplitude of the drain signals at D1 and D2 has not reached the full maximum yet.

ISO7741T-Q1 devices support soft-start feature. Upon power up, the gate drive of the output power-MOSFET is gradually increased over a period of time from 0V to V_{CC1} . Soft-start prevents high inrush current from V_{CC1} while charging large secondary side decoupling capacitors, and also prevents overshoot in secondary voltage during power-up.

7.4.3 Operating Mode

When the device supply has reached the nominal value $\pm 10\%$ the oscillator is fully operating. However variations over supply voltage and operating temperature can vary the switching frequencies at D1 and D2.

7.4.4 Spread Spectrum Clocking

Radiated emissions is an important concern in high current switching power supplies. ISO7741T transformer driver addresses this by modulating the internal clock in such a way that the emitting energy is spread over multiple frequency bins. This Spread Spectrum clocking feature greatly improves the emissions performance of the entire power supply block and hence relieves the system designer from one major concern in isolated power supply design.

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The device is a high-performance, quad-channel digital isolator with integrated high efficiency transformer driver. Typically digital isolators require two power supplies isolated from each other to power up both sides of device. Due to integrated transformer driver, with external transformer, the isolated supply can be generated to power the isolated side of the device and peripherals on isolated side, thus saving board space without compromising on efficiency. The voltage range is from 2.25V to 5.5V for both supplies, V_{CC1} and V_{CC2} . When designing with digital isolators, keep in mind that because of the single-ended design structure, digital isolators do not conform to any specific interface standard and are only intended for isolating single-ended CMOS or TTL digital signal lines. The isolator is typically placed between the data controller (that is, μC or UART), and a data converter or a line transceiver, regardless of the interface type or standard.

The device is designed for applications that have limited board space and desire more integration. The device is also designed for very high voltage applications with power transformer supporting high voltage isolation.

8.2 Typical Application

The following application circuit is shown for a 3.3V input supply commonly taken from the local, regulated microcontroller supply for SPI isolation. For 5V input voltages requiring different turn ratios refer to the transformer manufacturers and the web sites listed in [Transformer Manufacturers](#)

Table 8-1. Transformer Manufacturers

MANUFACTURER	MORE INFORMATION
Coilcraft Inc.	http://www.coilcraft.com
Halo-Electronics Inc.	http://www.haloelectronics.com
Murata Power Solutions	http://www.murata-ps.com
Würth Electronics Midcom Inc	http://www.midcom-inc.com

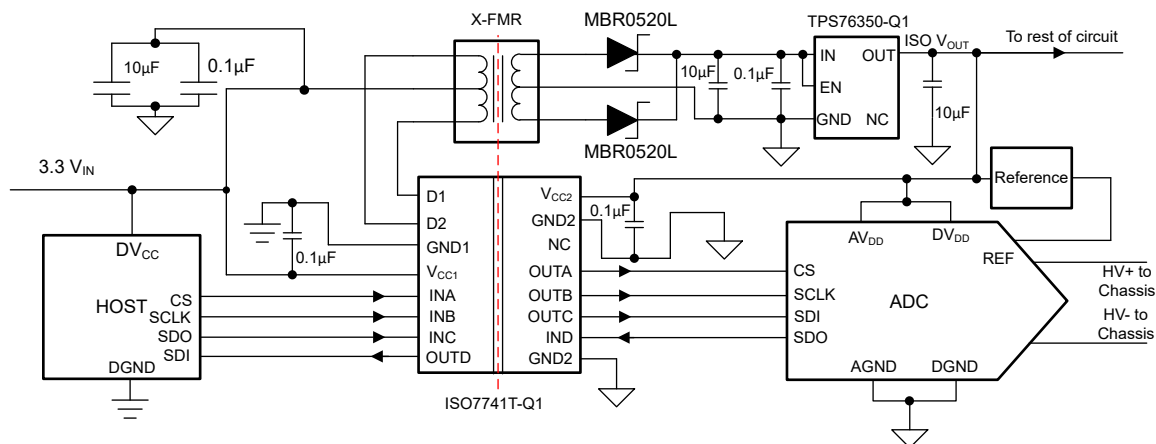


Figure 8-1. Isolated Power and SPI for ADC Sensing Application With ISO7741T-Q1

8.2.1 Design Requirements

To design with these devices, use the parameters listed in [Design Parameters](#).

Table 8-2. Design Parameters

PARAMETER	VALUE
Supply voltage, V_{CC1}	2.25V to 5.5V
Decoupling capacitor between V_{CC1} and GND1	0.1 μ F + 1 μ F to 10 μ F
Decoupling capacitor from V_{CC2} and GND2	0.1 μ F + 1 μ F to 10 μ F

8.2.2 Detailed Design Procedure

The following recommendations on components selection focus on the design of an efficient push-pull converter with high current drive capability. Contrary to popular belief, the output voltage of the unregulated converter output drops significantly over a wide range in load current. The characteristic curve in [Figure 8-2](#) and [Figure 8-3](#) for example, shows that the difference between V_{OUT} at minimum load and V_{OUT} at maximum load exceeds a transceiver supply range. Therefore, to provide a stable, load independent supply while maintaining maximum possible efficiency the implementation of a low dropout regulator (LDO) is strongly advised.

8.2.2.1 Drive Capability

The transformer driver is designed for low-power push-pull converters with input and output voltages in the range of 2.25V to 5.5V. While converter designs with higher output voltages are possible, care must be taken that higher turns ratios do not lead to primary currents that exceed the specified current limits of the device.

8.2.2.2 LDO Selection

The minimum requirements for a suitable low dropout regulator are:

- The current drive capability must slightly exceed the specified load current of the application to prevent the LDO from dropping out of regulation. Therefore, for a load current of 550mA, choose a 600mA to 700mA LDO. While regulators with higher drive capabilities are acceptable, these regulators also typically possess higher dropout voltages that reduce the overall converter efficiency.
- The internal dropout voltage, V_{DO} , at the specified load current must be as low as possible to maintain efficiency. For a low-cost 700mA LDO, a V_{DO} of 600mV at 700mA is common. Be aware; however, that this lower value is typically specified at room temperature and can increase by a factor of 2 over temperature, which in turn raises the required minimum input voltage.
- The required minimum input voltage preventing the regulator from dropping out of line regulation is given with:

$$V_{I-min} = V_{DO-max} + V_{O-max} \quad (1)$$

To determine V_I for worst-case condition, the user must take the maximum values for V_{DO} and V_O specified in the LDO data sheet for rated output current (that is, 600mA) and add these values together. The user must also specify that the output voltage of the push-pull rectifier at the specified load current is equal or higher than V_{I-min} . If the output voltage is not, the LDO loses line-regulation and any variations at the input passes straight through to the output. Hence, below V_{I-min} the output voltage follows the input and the regulator behaves like a simple conductor.

- The maximum regulator input voltage must be higher than the rectifier output under no-load. Under this condition there is no secondary current reflected back to the primary, thus making the voltage drop across R_{DS-on} negligible and allowing the entire converter input voltage to drop across the primary. At this point, the secondary reaches the maximum voltage of

$$V_{S-max} = V_{IN-max} \times n \quad (2)$$

with V_{IN-max} as the maximum converter input voltage and n as the transformer turns ratio. Thus to prevent the LDO from damage the maximum regulator input voltage must be higher than V_{S-max} . [Table 8-3](#) lists the maximum secondary voltages for various turns ratios commonly applied in push-pull converters.

Table 8-3. Required Maximum LDO Input Voltages for Various Push-Pull Configurations

PUSH-PULL CONVERTER				LDO
CONFIGURATION	V_{IN-max} [V]	TURNS-RATIO	V_{S-max} [V]	V_{I-max} [V]
$3.3V_{IN}$ to $3.3V_{OUT}$	3.6	$1.5 \pm 3\%$	5.6	6 to 10
$3.3V_{IN}$ to $5V_{OUT}$	3.6	$2.2 \pm 3\%$	8.2	10
$5V_{IN}$ to $5V_{OUT}$	5.5	$1.5 \pm 3\%$	8.5	10

8.2.2.3 Diode Selection

A rectifier diode must always possess low-forward voltage to provide as much voltage to the converter output as possible. When used in high-frequency switching applications, such as the ISO7741Tx-Q1 however, the diode must also possess a short recovery time. Schottky diodes meet both requirements and are therefore strongly recommended in push-pull converter designs. A good choice for low-volt applications and ambient temperatures of up to 85°C is the low-cost Schottky rectifier MBR0520L with a typical forward voltage of 275mV at 100mA forward current. For higher output voltages such as $\pm 10V$ and above use the MBR0530 which provides a higher DC blocking voltage of 30V.

Lab measurements have shown that at temperatures higher than 100°C the leakage currents of the above Schottky diodes increase significantly. These conditions can cause thermal runaway leading to the collapse of the rectifier output voltage. Therefore, for ambient temperatures higher than 85°C use low-leakage Schottky diodes, such as RB168MM-40.

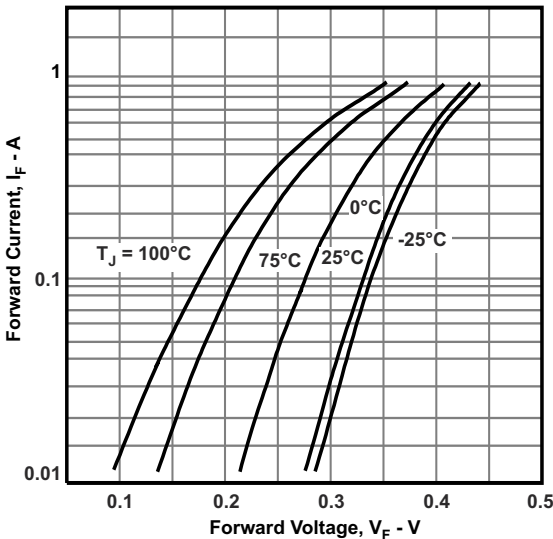


Figure 8-2. Diode Forward Characteristics for MBR0520L

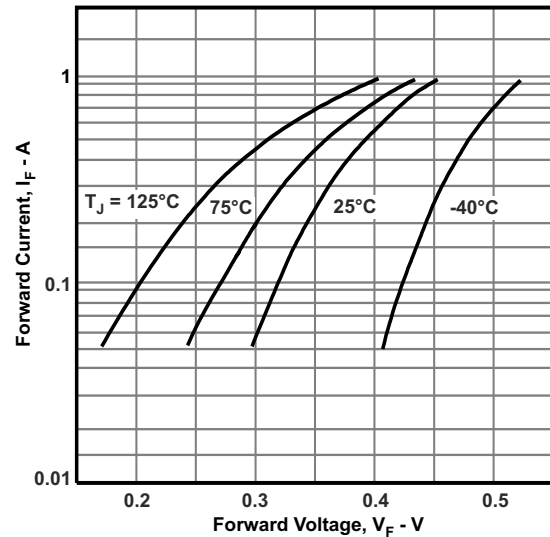


Figure 8-3. Diode Forward Characteristics for MBR0530

8.2.2.4 Capacitor Selection

As with all high speed CMOS ICs, the device requires a bypass capacitor in the range of 10nF to 100nF.

The input bulk capacitor at the center-tap of the primary supports large currents into the primary during the fast switching transients. For minimum ripple make this capacitor 1 μ F to 10 μ F. In a 2-layer PCB design with a dedicated ground plane, place this capacitor close to the primary center-tap to minimize trace inductance. In a 4-layer board design with low-inductance reference planes for ground and V_{CC} , the capacitor can be placed at the supply entrance of the board. To provide low-inductance paths use two vias in parallel for each connection to a reference plane or to the primary center-tap.

The bulk capacitor at the rectifier output smooths the output voltage. Make this capacitor 1 μ F to 10 μ F.

The small capacitor at the regulator input is not necessarily required. However, good analog design practice suggests, using a small value of 47nF to 100nF improves the transient response and noise rejection of the regulator.

The LDO output capacitor buffers the regulated output for the subsequent isolator and transceiver circuitry. The choice of output capacitor depends on the LDO stability requirements specified in the data sheet. However, in most cases, a low-ESR ceramic capacitor in the range of 4.7μF to 10μF satisfies these requirements.

8.2.2.5 Transformer Selection

8.2.2.5.1 V-t Product Calculation

To prevent a transformer from saturation, the V-t product must be greater than the maximum V-t product applied by the device. The maximum voltage delivered by the device is the nominal converter input plus 10%. The maximum time this voltage is applied to the primary is half the period of the lowest frequency at the specified input voltage. Therefore, the minimum V-t product for the transformer is determined through:

$$V_{t_{\min}} \geq V_{IN - \max} \times \frac{T_{\max}}{2} = \frac{V_{IN - \max}}{2 \times f_{\min}} \quad (3)$$

Taking an example of $f_{\min}$ as 138kHz for ISO7741TA-Q1 and 363kHz for ISO7741TB-Q1 with a 5V supply, the following equations yield the minimum V-t products of:

$$V_{t_{\min}} \geq \frac{5.5V}{2 \times 138kHz} = 20V\mu s \text{ for ISO7741TA-Q1 applications, and}$$

$$V_{t_{\min}} \geq \frac{5.5V}{2 \times 363kHz} = 7.6V\mu s \text{ for ISO7741TB-Q1 applications}$$

Common V-t values for low-power center-tapped transformers range from 22Vμs to 150Vμs with typical footprints of 10mm x 12mm. However, transformers specifically designed for PCMCIA applications provide as little as 11Vμs and come with a significantly reduced footprint of 6mm x 6mm only.

While Vt-wise all of these transformers can be driven by the device, other important factors such as isolation voltage, transformer wattage, and turns ratio must be considered before making the final decision.

8.2.2.5.2 Turns Ratio Estimate

Assuming that the rectifier diodes and linear regulator are selected and that the transformer selected must have a V-t product of at least 11Vμs. However, before searching the manufacturer web sites for a suitable transformer, the user still needs to know the minimum turns ratio that allows the push-pull converter to operate flawlessly over the specified current and temperature range. This minimum transformation ratio is expressed through the ratio of minimum secondary to minimum primary voltage multiplied by a correction factor that takes the transformer typical efficiency of 97% into account:

$$V_{P - \min} = V_{IN - \min} - V_{DS - \max} \quad (4)$$

$V_{S - \min}$ must be large enough to allow for a maximum voltage drop, $V_{F - \max}$, across the rectifier diode and still provide sufficient input voltage for the regulator to remain in regulation. From the [LDO Selection](#) section, this minimum input voltage is known and by adding $V_{F - \max}$ gives the minimum secondary voltage with:

$$V_{S - \min} = V_{F - \max} + V_{DO - \max} + V_{O - \max} \quad (5)$$

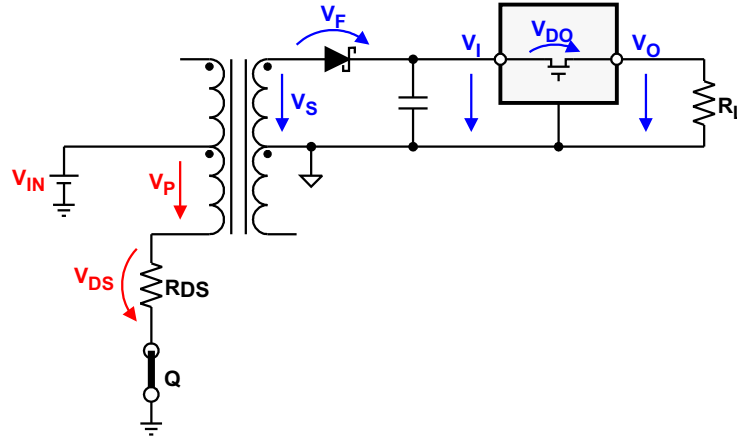


Figure 8-4. Establishing the Required Minimum Turns Ratio Through $N_{\min} = 1.031 \times V_{S-\min} / V_{P-\min}$

Then calculating the available minimum primary voltage, $V_{P-\min}$, involves subtracting the maximum possible drain-source voltage of the device, $V_{DS-\max}$, from the minimum converter input voltage $V_{IN-\min}$:

$$V_{P-\min} = V_{IN-\min} - V_{DS-\max} \quad (6)$$

$V_{DS-\max}$ however, is the product of the maximum $R_{DS(on)}$ and I_D values for a given supply specified in the data sheet:

$$V_{DS-\max} = R_{DS-\max} \times I_{D-\max} \quad (7)$$

Then inserting Equation 7 into Equation 6 yields:

$$V_{P-\min} = V_{IN-\min} - R_{DS-\max} \times I_{D-\max} \quad (8)$$

and inserting Equation 8 and Equation 5 into Equation 4 provides the minimum turns ration with:

$$n_{\min} = 1.031 \times \frac{V_{F-\max} + V_{DO-\max} + V_{O-\max}}{V_{IN-\min} - R_{DS-\max} \times I_{D-\max}} \quad (9)$$

Example:

For a 3.3V_{IN} to 5V_{OUT} converter using the rectifier diode MBR0520L and the 5V LDO, the data sheet values taken for a load current of 600mA and a maximum temperature of 85°C are $V_{F-\max} = 0.2V$, $V_{DO-\max} = 0.5V$, and $V_{O-\max} = 5.1V$.

Then assuming that the converter input voltage is taken from a 3.3V controller supply with a maximum $\pm 2\%$ accuracy makes $V_{IN-\min} = 3.234V$. Finally the maximum values for drain-source resistance and drain current at 3.3V are taken from the data sheet with $R_{DS-\max} = 0.45\Omega$ and $I_{D-\max} = 700mA$.

Inserting the values above into Equation 10 yields a minimum turns ratio of:

$$n_{\min} = 1.031 \times \frac{0.2V + 0.5V + 5.1V}{3.234V - 0.45\Omega \times 700mA} \quad (10)$$

Most commercially available transformers for 3V to 5V push-pull converters offer turns ratios between 2.0 and 2.3 with a common tolerance of $\pm 3\%$.

8.2.2.5.3 Recommended Transformers

Depending on the application, use the minimum configuration in [Figure 8-5](#) or standard configuration in [Figure 8-6](#).

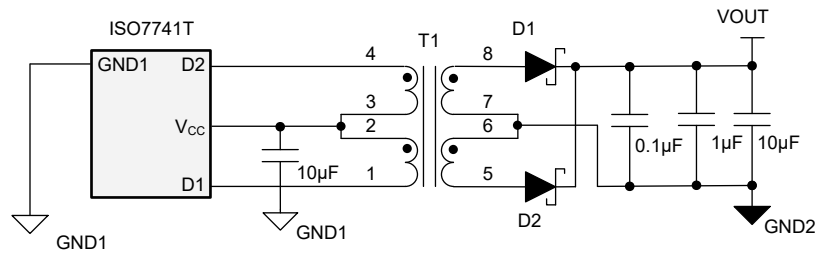


Figure 8-5. Unregulated Output for Low-Current Loads With Wide Supply Range

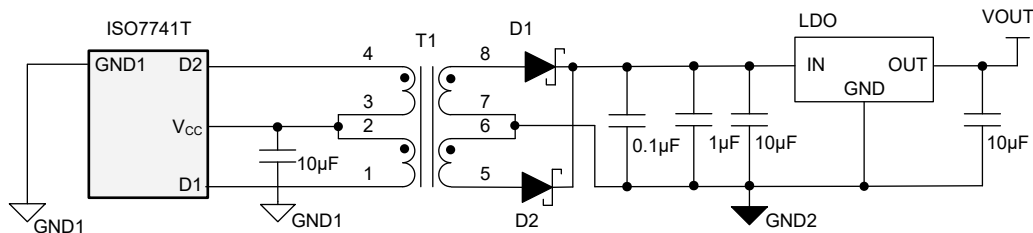


Figure 8-6. Regulated Output for Stable Supplies and High Current Loads

The Würth Electronics Midcom isolation transformers in [Table 8-4](#) are optimized designs for the device, providing high efficiency and small form factor at low-cost.

The 1:1.1 and 1:1.7 turns-ratios are designed for logic applications with wide supply rails and low load currents. These applications operate without LDO, thus achieving further cost-reduction.

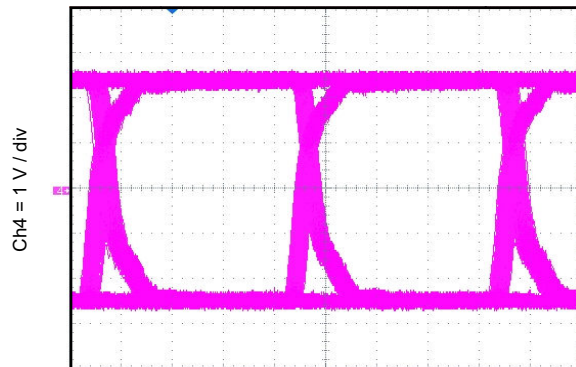
Table 8-4. Recommended Isolation Transformers Optimized for the Device

Turns Ratio	V × T (Vμs)	Isolation (V _{RMS})	Dimensions (mm)	Application	LDO ⁽¹⁾	Order No.	Manufacturer	
1:1.1 ±2%	7	2500	6.73 x 10.05 x 4.19	3.3V → 3.3V, 100mA, ISO7741TB-Q1	No	760390011	Würth Electronics / Midcom	
1:1.1 ±2%	11			5V → 5V, 100mA, ISO7741TB-Q1		760390012		
1:1.7 ±2%				3.3V → 5V, 100mA, ISO7741TB-Q1		760390013		
1:1.3 ±2%				3.3V → 3.3V, 100mA, ISO7741TB-Q1	760390014			
1:1.3 ±2%				5V → 5V, 100mA, ISO7741TB-Q1	760390014			
1:2.1 ±2%				3.3V → 5V, 100mA, ISO7741TB-Q1	760390015			
1.23:1 ±2%				5V → 3.3V, 100mA, ISO7741TB-Q1	750313710			
1:1.7 ±2%	8.9		8.3 x 12.6 x 4.1	3.3V → 3.3V, 1A, ISO7741TB-Q1	Yes	750316028		
1:2.1 ±2%				3.3V → 5V, 1A, ISO7741TB-Q1		750316029		
1.3:1 ±2%	5V → 3.3V, 1A, ISO7741TB-Q1			750316030				
1:1.1 ±2%	8.6			3.3V → 3.3V, 1A, ISO7741TB-Q1 5V → 5V, 1A, ISO7741TB-Q1		No		750315371
								750313734
1:1.1 ±2%	11	9.14 x 12.7 x 7.37		5V → 5V, 100mA, ISO7741TB-Q1		No		750313734
1:1.1 ±2%			3.3V → 5V, 100mA, ISO7741TB-Q1	750313769				
1:1.7 ±2%			3.3V → 3.3V, 100mA, ISO7741TB-Q1 5V → 5V, 100mA, ISO7741TB-Q1	Yes	750313638			
1:1.3 ±2%			3.3V → 5V, 100mA, ISO7741TB-Q1		750313626			
1:2.1 ±2%			5V → 3.3V, 100mA, ISO7741TB-Q1		No	750313638		
1.3:1 ±2%			41	12.32 x 15.41 x 11.05	3.3V → 3.3V, 1A, ISO7741TA-Q1	Yes		750316031
1:1.75 ±2%	3.3V → 5V, 1A, ISO7741TA-Q1	No			750316032			
1:2 ±2%	5.0V → 3.3V, 1A, ISO7741TA-Q1				750316033			
1.3:1 ±2%	3.3V → 3.3V, 1A, ISO7741TA-Q1 5V → 5V, 1A, ISO7741TA-Q1				750315240			
1:1.1 ±2%	23	12.32 x 15.41 x 11.89			5V → 17.5V, 100mA,ISO7741TB-Q1	No		750342879
1:3.5 ±2%					9.17 x 12.7 x 7.62			5V → 19.5V, 100mA,ISO7741TB-Q1
1:3.9 ±2%			9.5	8.3 x 12.6 x 4.1	5V → 18.75V, 100mA,ISO7741TB-Q1			78931812518
1:3.75 ±2%	5V → 23.75V, 100mA,ISO7741TB-Q1	78931812523						
1:4.75 ±2%	5V → 12.5V, 200mA,ISO7741TB-Q1	78931812512						
1:2.5 ±2%	5V → 15.65V, 200mA,ISO7741TB-Q1	78931812515						
1:3.13 ±2%	16	6000	9.14 x 12.7 x 7.62	5V → 17.5V, 100mA,ISO7741TB-Q1	No	750320340		
1:3.5 ±2%								
1:1.3 ±3%	11	5000	10.4 x 12.2 x 6.1	3.3V → 3.3V, 300mA, ISO7741TB-Q1 5V → 5V, 300mA, ISO7741TB-Q1	No	HCT-SM-1.3-8-2		Bourns
1:1.5 ±3%	34.4	2500	10 x 12.07 x 5.97	3.3V → 3.3V, 1A, ISO7741TB-Q1 5V → 5V, 1A, ISO7741TA/B-Q1	Yes	DA2303-AL	Coilcraft	
1:2.2 ±3%	21.5	2500	10 x 12.07 x 5.97	3.3V → 5V, 1A, ISO7741TA/B-Q1		DA2304-AL		

(1) For configurations with LDO, a higher voltage than the required output voltage is generated, to allow for LDO drop-out. Figures show the voltage and efficiency at the LDO input.

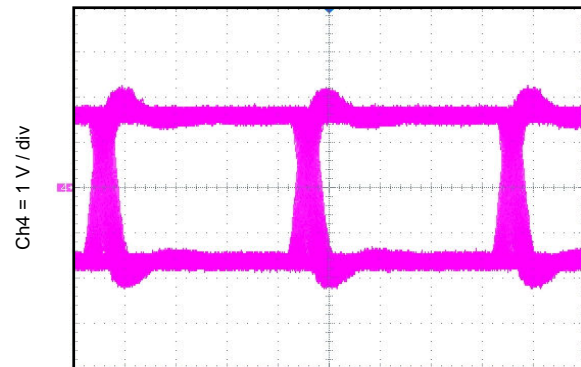
8.2.3 Application Curve

The following typical eye diagrams of the device indicates low jitter and wide open eye at the maximum data rate of 100Mbps.



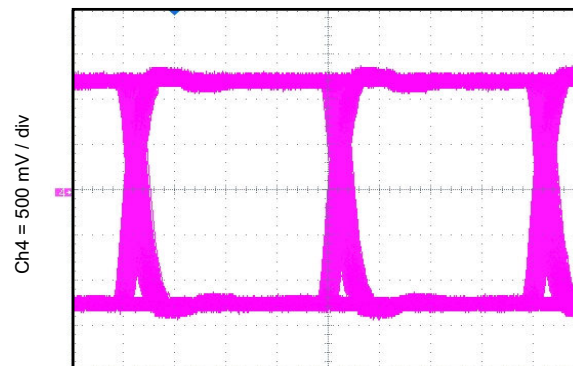
Time = 2.5 ns / div

Figure 8-7. Eye Diagram at 100Mbps PRBS $2^{16} - 1$, 5V and 25°C



Time = 2.5 ns / div

Figure 8-8. Eye Diagram at 100Mbps PRBS $2^{16} - 1$, 3.3V and 25°C



Time = 2.5 ns / div

Figure 8-9. Eye Diagram at 100Mbps PRBS $2^{16} - 1$, 2.5V and 25°C

8.2.3.1 Insulation Lifetime

Insulation lifetime projection data is collected by using industry-standard Time Dependent Dielectric Breakdown (TDDB) test method. In this test, all pins on each side of the barrier are tied together creating a two-terminal device and high voltage applied between the two sides; See [Figure 8-10](#) for TDDB test setup. The insulation breakdown data is collected at various high voltages switching at 60Hz over temperature. For reinforced insulation, VDE standard requires the use of TDDB projection line with failure rate of less than 1 part per million (ppm). Even though the expected minimum insulation lifetime is 20 years at the specified working isolation voltage, VDE reinforced certification requires additional safety margin of 20% for working voltage and 50% for lifetime which translates into minimum required insulation lifetime of 30 years at a working voltage that's 20% higher than the specified value.

[Figure 8-11](#) shows the intrinsic capability of the isolation barrier to withstand high voltage stress over its lifetime. Based on the TDDB data, the intrinsic capability of the insulation is 1500V_{RMS} with a lifetime of 36 years. Other factors, such as package size, pollution degree, material group, and more can further limit the working voltage of the component. The working voltage of DW-16 package is specified up to 1500V_{RMS}. At the lower working voltages, the corresponding insulation lifetime is much longer than 36 years.

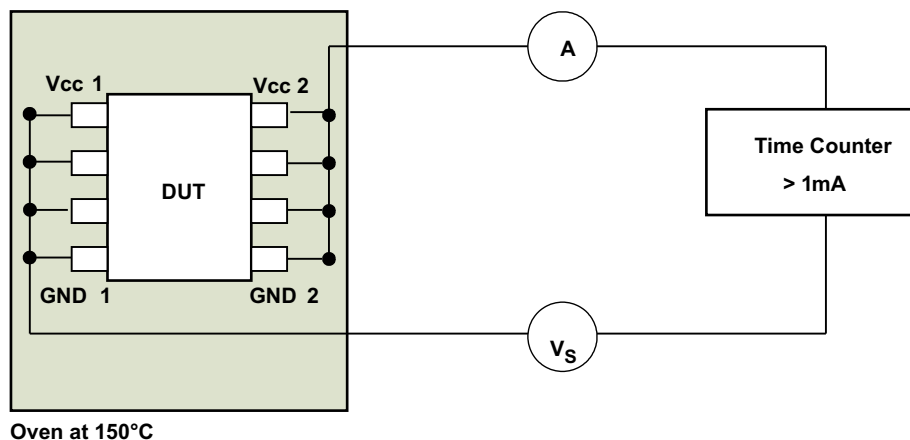


Figure 8-10. Test Setup for Insulation Lifetime Measurement

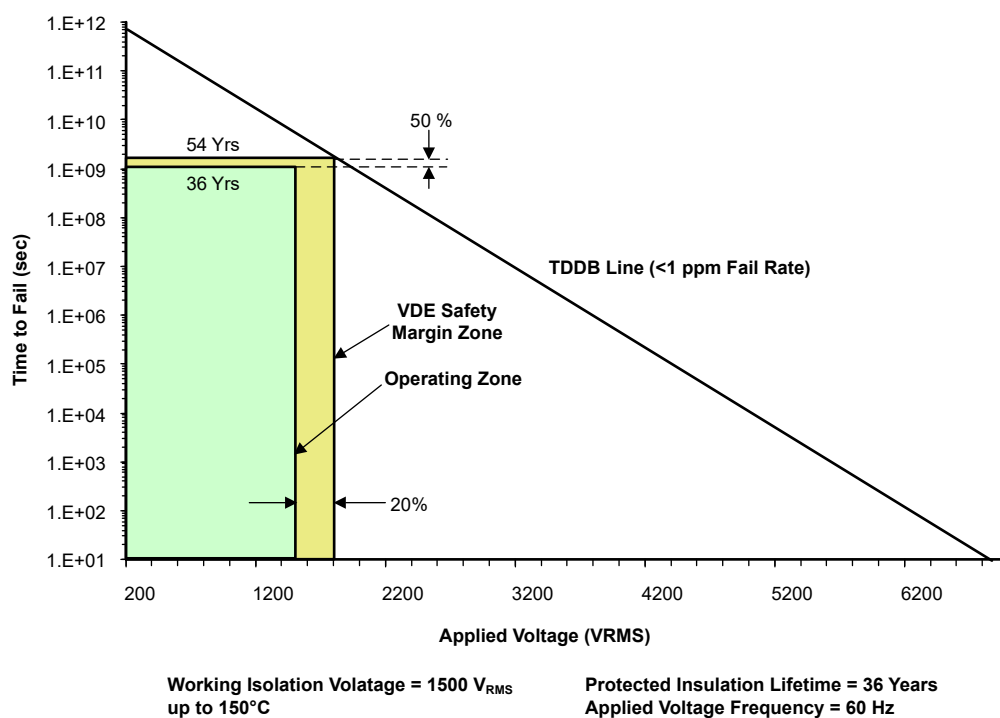


Figure 8-11. Insulation Lifetime Projection Data

8.2.4 System Examples

8.2.4.1 Higher Output Voltage Designs

The device can drive push-pull converters that provide high output voltages of up to 30V, or bipolar outputs of up to $\pm 15V$. Using commercially available center-tapped transformers, with relatively low turns ratios of 0.8 to 5, requires different rectifier topologies to achieve high output voltages. Figure 8-12 to Figure 8-14 show some of these topologies together with the respective open-circuit output voltages.

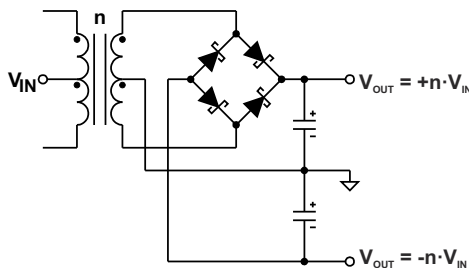


Figure 8-12. Bridge Rectifier With Center-Tapped Secondary Enables Bipolar Outputs

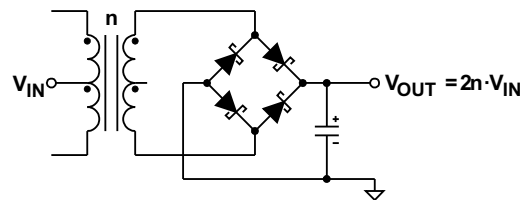


Figure 8-13. Bridge Rectifier Without Center-Tapped Secondary Performs Voltage Doubling

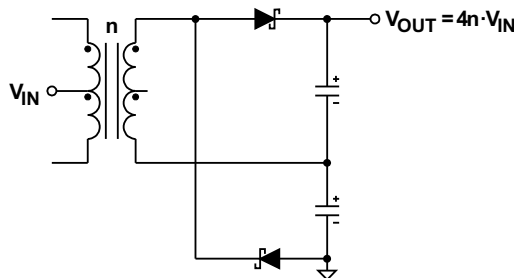


Figure 8-14. Half-Wave Rectifier Without Centered Ground and Center-Tapped Secondary Performs Voltage Doubling Twice, Hence Quadrupling V_{IN}

8.3 Power Supply Recommendations

The device is designed to operate in an input voltage supply range of 2.5V to 5V nominal. This input supply must be regulated within $\pm 10\%$. To help ensure reliable operation at data rates and supply voltages, a 0.1 μ F bypass capacitor is recommended at the input and output supply pins (V_{CC1} and V_{CC2}). The capacitors should be placed as close to the supply pins as possible. A 10 μ F capacitor should be connected close to the transformer center-tap pin for reliable power transfer. The isolated power is generated for the secondary-side with help of a transformer driver as shown in the [application diagram](#).

8.4 Layout

8.4.1 Layout Guidelines

A minimum of four layers is required to accomplish a low EMI PCB design (see Figure 8-15). Layer stacking must be in the following order (top-to-bottom): high-speed signal layer, ground plane, power plane and low-frequency signal layer.

- Routing the high-speed traces on the top layer avoids the use of vias (and the introduction of the inductances) and allows for clean interconnects between the isolator and the transmitter and receiver circuits of the data link.
- Placing a solid ground plane next to the high-speed signal layer establishes controlled impedance for transmission line interconnects and provides an excellent low-inductance path for the return current flow.
- Placing the power plane next to the ground plane creates additional high-frequency bypass capacitance of approximately 100pF/inch².
- Routing the slower speed control signals on the bottom layer allows for greater flexibility as these signal links usually have margin to tolerate discontinuities such as vias.

- A low-ESR ceramic bypass-capacitor should be connected between the V_{CC} and GND pins for both the supply. The recommended capacitor value can range from 0.1 μ F to 10 μ F. The capacitor must have a voltage rating of 10V minimum and a X5R or X7R dielectric. The optimum placement of decap is closest to the V_{CC} and GND pins.
- The connections between the device D1 and D2 pins and the transformer primary endings, and the connection of the device V_{CC1} pin and the transformer center-tap must be as close as possible for minimum trace inductance. And 10 μ F capacitor should be connected close to the transformer center-tap pin. Length matching of D1 and D2 traces will provide best performance in efficiency and EMI.
- The rectifier diodes should be Schottky diodes with low forward voltage in the 10mA to 100mA current range to maximize efficiency.

If an additional supply voltage plane or signal layer is needed, add a second power or ground plane system to the stack to keep the planes symmetrical. This design makes the stack mechanically stable and prevents warping. Also the power and ground plane of each power system can be placed closer together, thus increasing the high-frequency bypass capacitance significantly.

For detailed layout recommendations, refer to the [Digital Isolator Design Guide](#).

8.4.1.1 PCB Material

For digital circuit boards operating below 150Mbps, (or rise and fall times higher than 1ns), and trace lengths of up to 10 inches, use standard FR-4 UL94V-0 printed circuit boards. This PCB is preferred over cheaper alternatives due to the lower dielectric losses at high frequencies, less moisture absorption, greater strength and stiffness, and self-extinguishing flammability-characteristics.

8.4.2 Layout Example

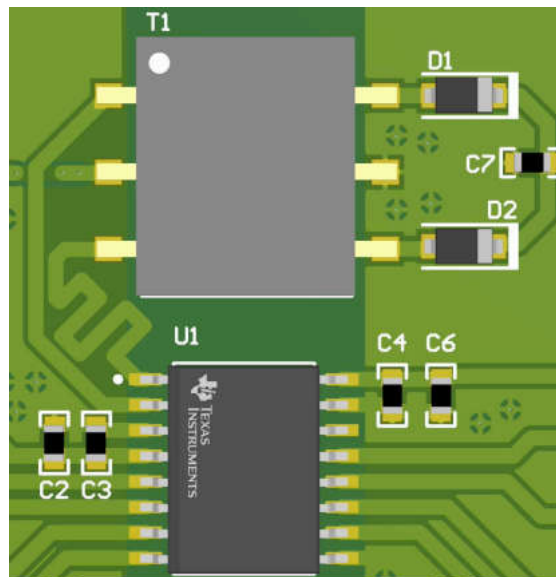


Figure 8-15. Example Layout

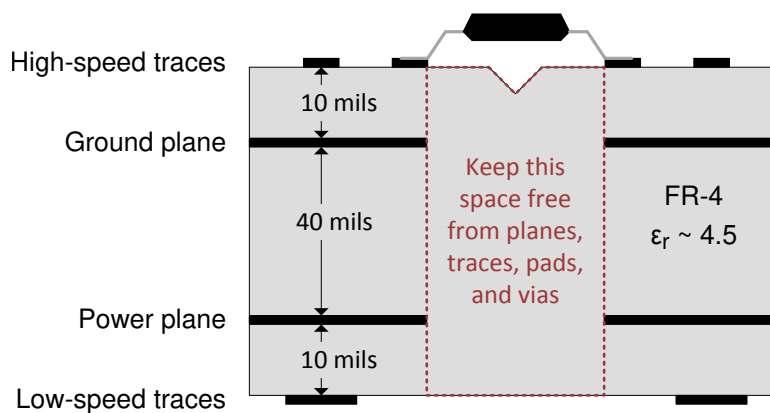


Figure 8-16. Example PCB Stackup

9 Device and Documentation Support

9.1 Documentation Support

9.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Digital Isolator Design Guide](#), application note
- Texas Instruments, [Isolation Glossary](#), application note
- Texas Instruments, [How to use isolation to improve ESD, EFT, and Surge immunity in industrial systems](#), application note

9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

9.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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9.4 Trademarks

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9.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

9.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

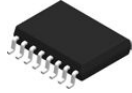
10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision * (April 2024) to Revision A (July 2024)	Page
• Updated the numbering format for tables, figures, and cross-references throughout the documents.....	1
• Changed minimum CMTI from 40 to 85 in all Electrical Characteristics tables.....	11

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical packaging and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

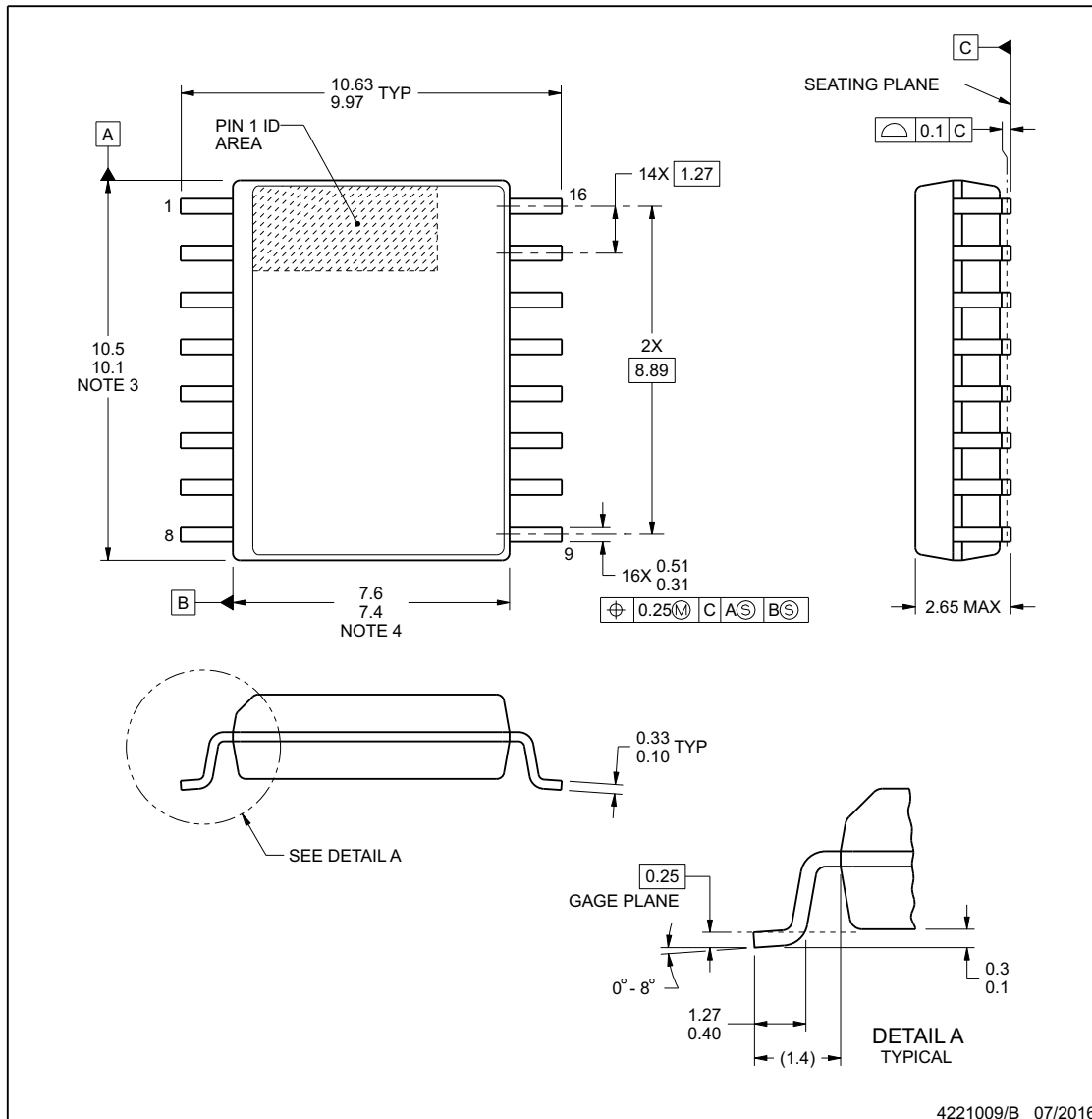


DW0016B

PACKAGE OUTLINE

SOIC - 2.65 mm max height

SOIC



NOTES:

1. All linear dimensions are in millimeters. Dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
5. Reference JEDEC registration MS-013.

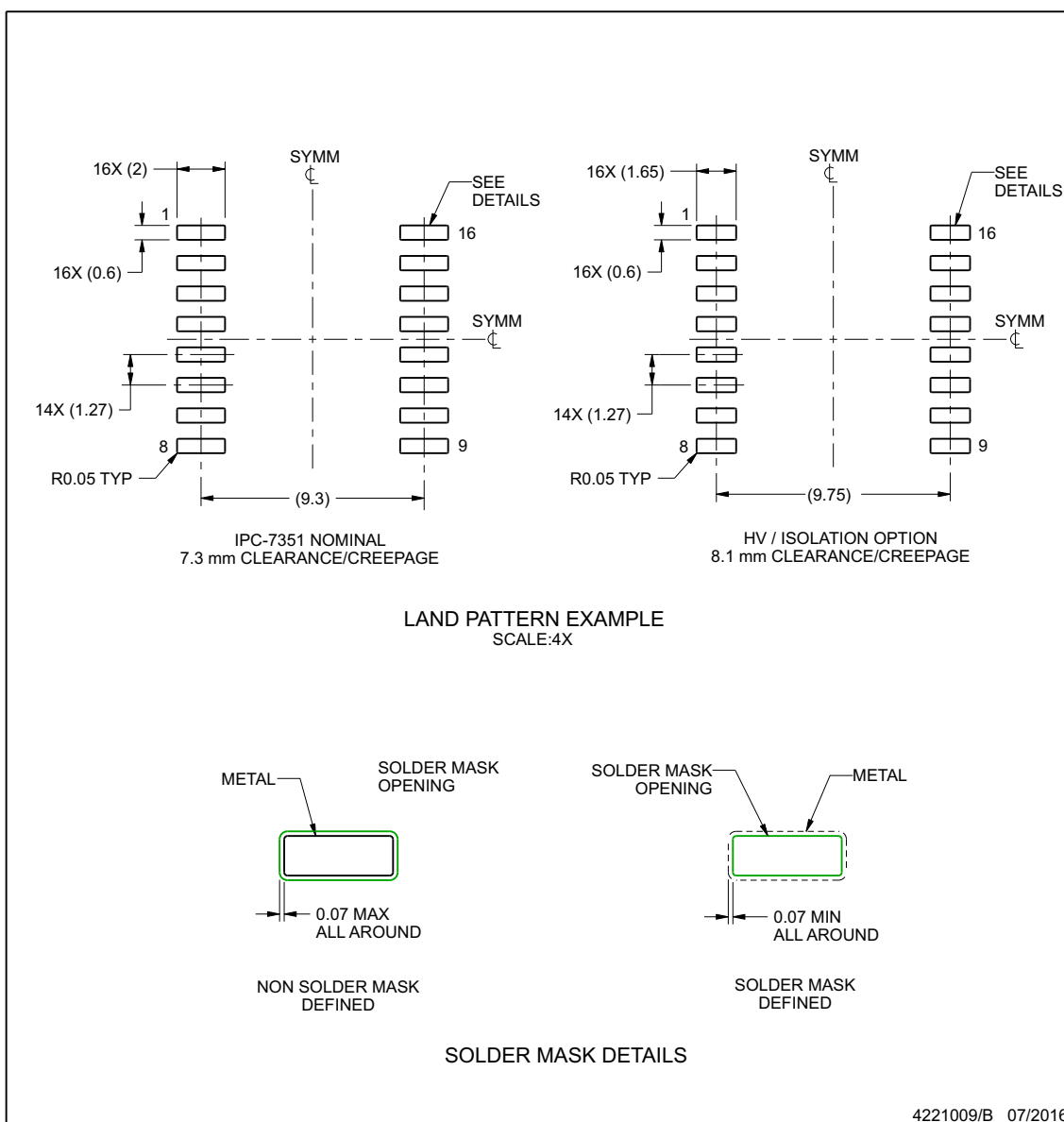
www.ti.com

EXAMPLE BOARD LAYOUT

DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

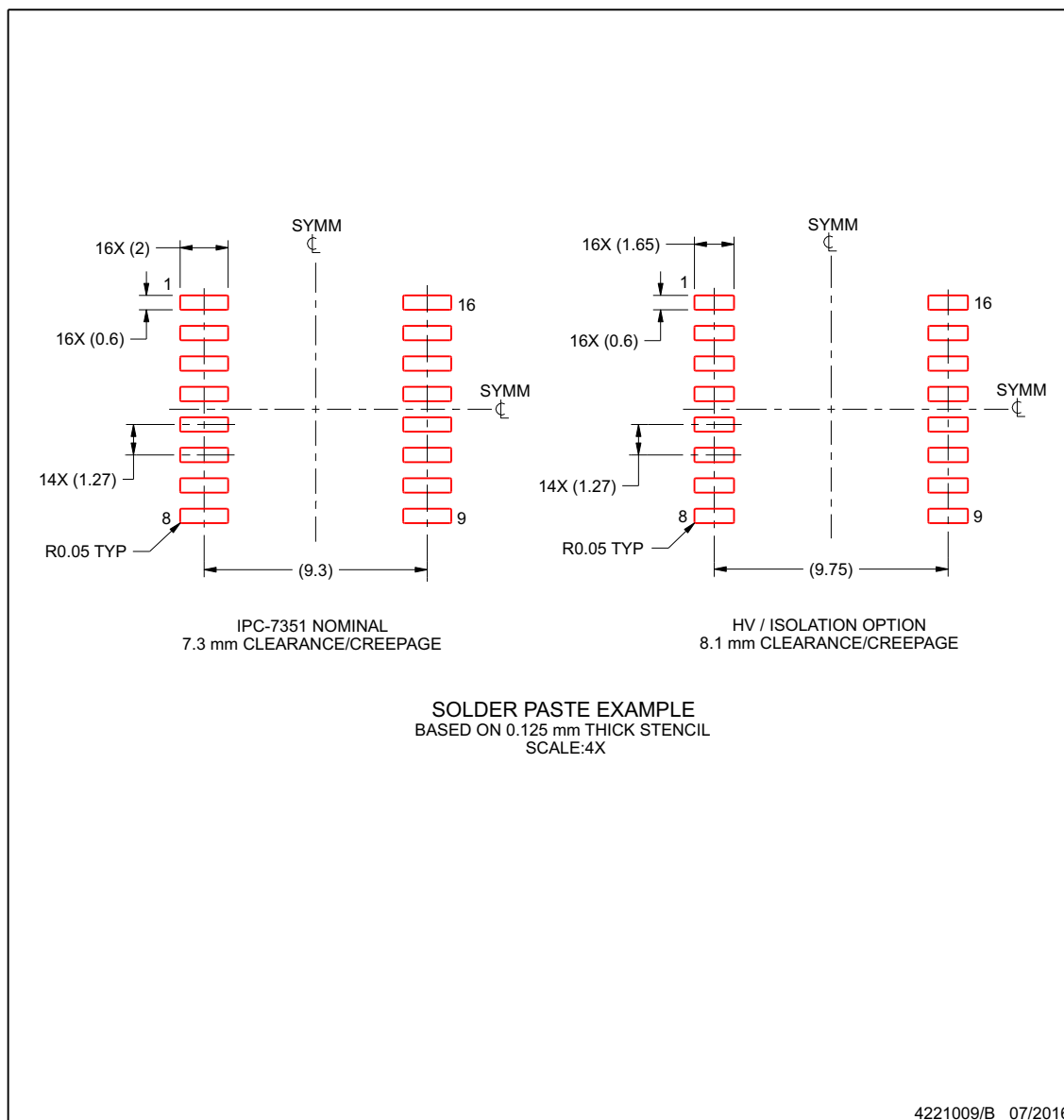
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EXAMPLE STENCIL DESIGN

DW0016B

SOIC - 2.65 mm max height

SOIC



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
ISO7741FTADWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7741FTA	Samples
ISO7741FTBDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7741FTB	Samples
ISO7741TADWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7741TA	Samples
ISO7741TBDWRQ1	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	ISO7741TB	Samples
XISO7741TADWRQ1	ACTIVE	SOIC	DW	16	2000	TBD	Call TI	Call TI	-40 to 125		Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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