

采用 SC70 封装、具有内部增益和滤波器的 3 V 视频放大器

查询样品: [OPA361-Q1](#)

特性

- 符合汽车应用要求
 - 出色的视频性能
 - 内部增益: **5.2V/V**
 - 支持电视检测
 - 与 **OMAP242x** 及达芬奇 (**DAVINCI™**) 处理器兼容
 - **2** 极重建滤波器
 - 输入范围包括接地
 - **DC** 耦合输入
 - 集成型电平转换器
 - **DC** 耦合输出⁽¹⁾
 - 无需输出电容器
 - 轨至轨输出
 - 低静态电流: **5.3mA**
 - 关断电流: **1.5μA**
 - 单电源: **2.5 V 至 3.3 V**
 - **SC70-6** 封装: **2.0 毫米 × 2.1 毫米**
 - 符合 **RoHS** 标准
- (1) 内部电路系统可避免输出饱和, 即便是输入视频信号提供 0 V 同步脉冲幅度也是如此。

说明

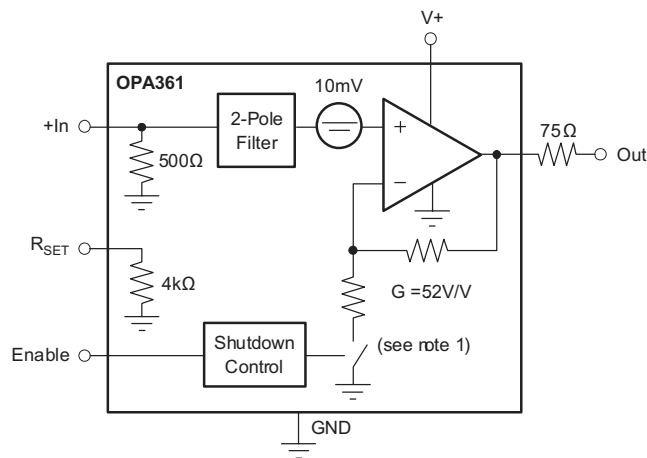
OPA361-Q1 高速放大器针对 3 V 便携式视频应用进行了优化。它经过精心设计, 与德州仪器 (TI) OMAP2420 及达芬奇处理器或支持 0.5 V_{PP} 视频输出的其它应用处理器中嵌入的视频编码器兼容。输入共模范围涵盖 GND, 这使视频数模转换器 (DAC) 能够与 OPA361-Q1 实现 DC 耦合。电视检测特性可通过协助实现视频传输的自动启动 / 停止来大幅简化最终用户接口。

5mV 内的 GND 输出摆幅, 250mV v+ 支持标准反向端接负载 (150Ω)。内部电平转换可防止输出出现 0V 输出饱和, 从而可在通用视频电路中防止同步脉冲削波。因此, OPA361-Q1 是 DC 耦合视频负载的理想选择。

OPA361-Q1 高度集成内部增益设置电阻器 ($G = 5.2V/V$) 与 2 极视频 DAC 重建滤波器, 针对空间狭小的应用进行了优化。

在关断模式下, 静态电流可降至 1.5μA 以下, 从而可显著降低功耗, 延长电池使用寿命。

OPA361-Q1 采用微型 2 毫米 × 2.1 毫米 SC70-6 封装。



(1) 在正常工作下启动时关闭; 关断时打开。



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

PRODUCTION DATA information is current as of publication date. Products conform to specifications per the terms of the Texas Instruments standard warranty. Production processing does not necessarily include testing of all parameters.

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English Data Sheet: [SBOS552](#)

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

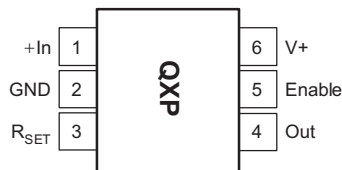
ORDERING INFORMATION⁽¹⁾

PRODUCT	PACKAGE	PACKAGE DESIGNATOR	PACKAGE MARKING
OPA361-Q1	SC70-6	DCK	QXP

- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this datasheet, or see the TI website at www.ti.com.

PIN CONFIGURATION

DCK PACKAGE
(TOP VIEW)



The location of pin 1 on the OPA361-Q1 is determined by orienting the package marking as shown in the diagram above.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

	VALUE	UNIT
Supply voltage, V+ to V–	+3.6	V
Signal input terminals	Voltage ⁽²⁾	–0.5 to (V+) + 0.5
	Current ⁽²⁾	±10
Output short-circuit through 75Ω to GND ⁽³⁾	Continuous	
Operating temperature	–40 to +125	°C
Storage temperature	–65 to +150	°C
Junction temperature	+160	°C

- (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.
- (3) Short-circuit to ground.

ELECTRICAL CHARACTERISTICS: $V_S = +2.5V$ to $+3.3V$

Boldface limits apply over the temperature range, $T_A = -40^{\circ}C$ to $+125^{\circ}C$.

At $T_A = +25^{\circ}C$, $R_L = 150\Omega$ connected to GND, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OFFSET LEVEL-SHIFT VOLTAGE						
V_{OLS}	Output Level-Shift Voltage ⁽¹⁾	$V_S = +2.8V$, $V_{IN} = GND$	-3	11	55	mV
	Over Temperature			20		mV
PSRR	vs Power Supply	$V_S = +2.5V$ to $+3.3V$		± 80		$\mu V/V$
INPUT VOLTAGE RANGE						
V_{CM}	Input Voltage Range ⁽²⁾	$V_S = 2.5V$	GND		0.42	V
		$V_S = 2.8V$	GND		0.48	V
		$V_S = 3.3V$	GND		0.55	V
R_{IN}	Input Resistance (+In)		450	510	550	Ω
R_{SET}	R_{SET} Resistance		3600	4070	4400	Ω
	Matching of R_{IN} and R_{SET}		2%	0.5%		
VOLTAGE GAIN						
		$\Delta V_{OUT}/\Delta V_{IN}$, $V_S = +2.5V$, $V_{INMIN} = 0V$, $V_{INMAX} = 0.42V$	5.06	5.17	5.28	V/V
		$\Delta V_{OUT}/\Delta V_{IN}$, $V_S = +2.8V$, $V_{INMIN} = 0V$, $V_{INMAX} = 0.48V$	5.06	5.17	5.28	V/V
		$\Delta V_{OUT}/\Delta V_{IN}$, $V_S = +3.3V$, $V_{INMIN} = 0V$, $V_{INMAX} = 0.55V$	5.06	5.17	5.28	V/V
FREQUENCY RESPONSE						
	Filter Response					
f-3dB	Cutoff Frequency			9		MHz
Normalized Gain:	$f_{IN} = 4.5MHz$	$V_O = 2V_{PP}$	-0.1			dB
	$f_{IN} = 27MHz$	$V_O = 2V_{PP}$	-18			dB
	$f_{IN} = 54MHz$	$V_O = 2V_{PP}$	-23			dB
	Differential Gain Error	$R_L = 150\Omega$		1.2%		
	Differential Phase Error	$R_L = 150\Omega$		1.6		degrees
	Group Delay Variation	100kHz, 4.5MHz		26		ns
SNR	Signal-to-Noise Ratio	100% White Signal		65		dB
OUTPUT						
	Positive Voltage Output Swing from Rail	$V_S = +2.8V$, $V_{IN} = 0.7V$, Ω to GND		130	250	mV
	Negative Voltage Output Swing from Rail	$V_S = +2.8V$, $V_{IN} = -0.05V$, $R_L = 150\Omega$ to GND		0.15	5	mV
	Positive Voltage Output Swing from Rail	$V_S = +2.8V$, $V_{IN} = 0.7V$, $R_L = 75\Omega$ to GND		260		mV
	Negative Voltage Output Swing from Rail	$V_S = +2.8V$, $V_{IN} = -0.05V$, $R_L = 75\Omega$ to GND		2		mV
	Output Leakage	$V_S = +2.8V$, Disabled, $V_O = 2V$		0.3	100	nA
I_O	Output Current ⁽³⁾	$V_S = +2.8V$		± 80		mA
POWER SUPPLY						
V_S	Specified Voltage Range		2.5		3.3	V
I_Q	Quiescent Current	$V_S = +2.8V$, Enabled, $I_O = 0$, $V_{OUT} = 1.4V$		5.3	7.5	mA
	Over Temperature	Specified Temperature Range			9	mA
ENABLE/SHUTDOWN FUNCTION						
	Disabled (logic-LOW threshold)		0		0.35	V
	Enabled (logic-HIGH threshold)		1.3		V_S	V
	Enable Time			1.5		μs
	Disable Time			50		ns
	Shutdown Current	$V_S = +2.8V$, Disabled		1.5	3	μA
TEMPERATURE RANGE						
	Specified/Operating Range		-40		+125	$^{\circ}C$
	Storage Range		-65		+150	$^{\circ}C$
θ_{JA}	Thermal Resistance					
	SC70			250		$^{\circ}C/W$

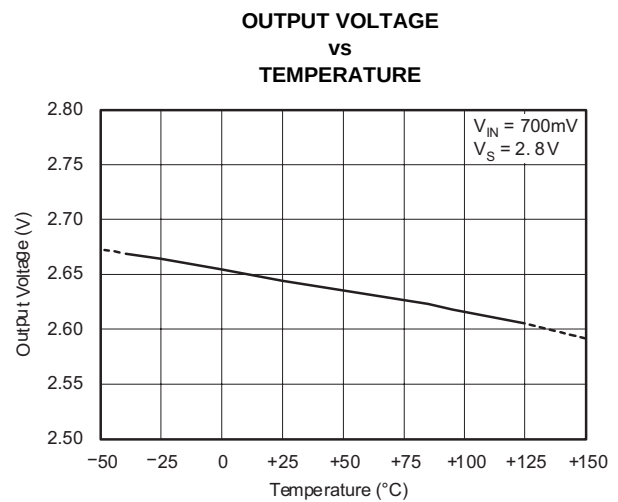
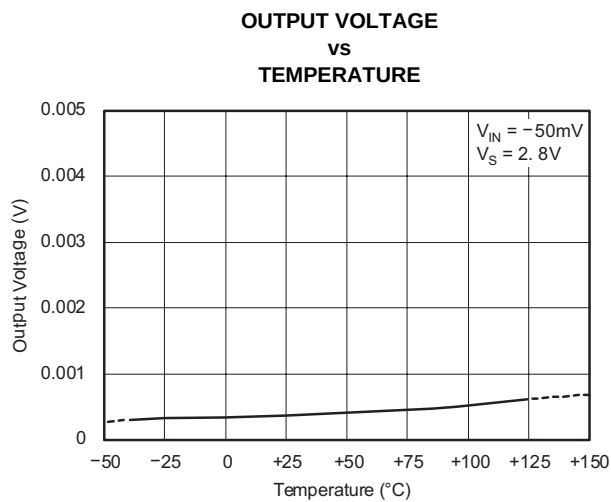
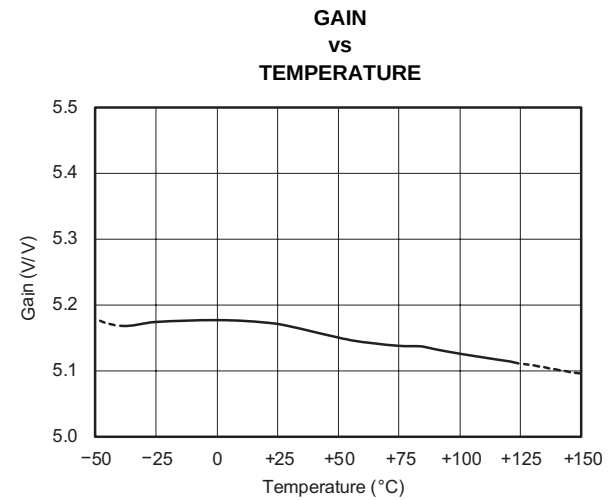
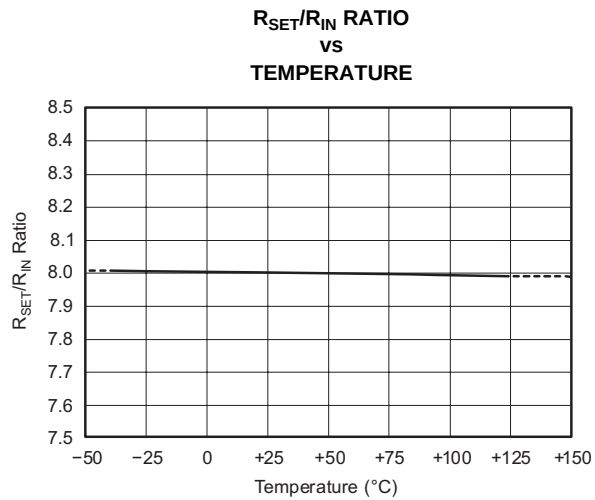
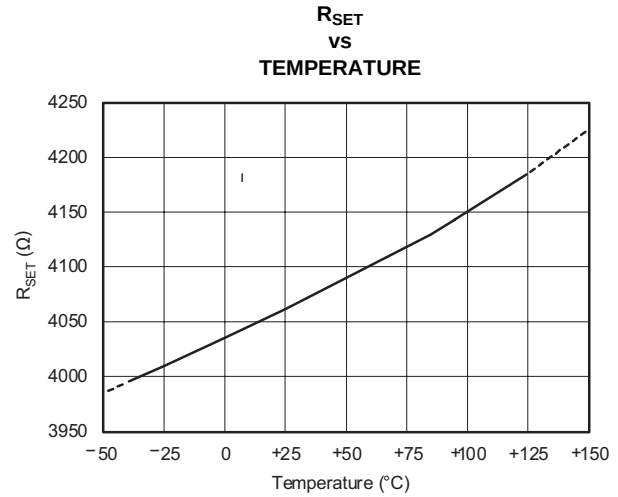
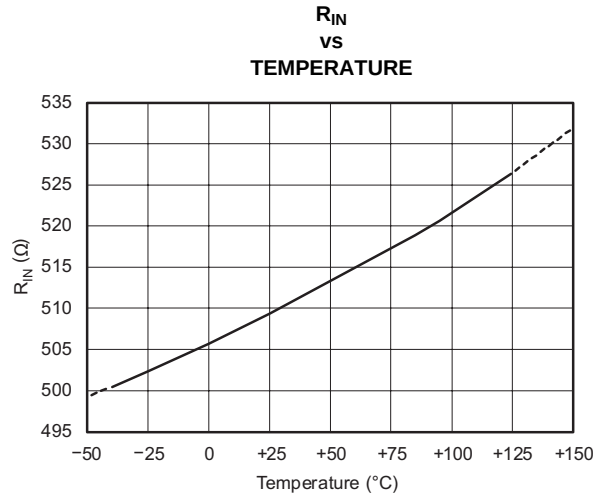
(1) Output referred.

(2) Limited by output swing and internal $G = 5.2V/V$.

(3) See typical characteristics *Output Voltage Swing vs Output Current*.

TYPICAL CHARACTERISTICS: $V_S = 2.8\text{ V}$

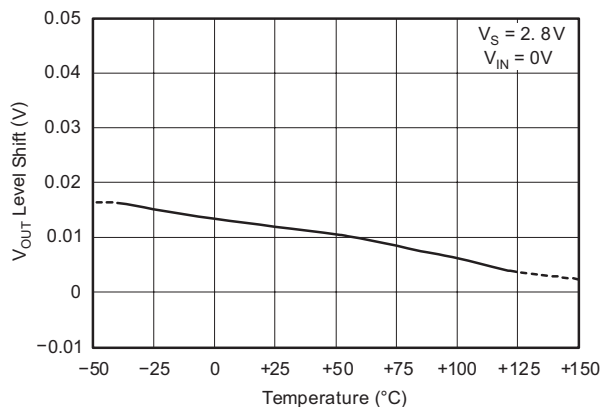
At $T_A = +25^\circ\text{C}$ and $R_L = 150$, unless otherwise noted.



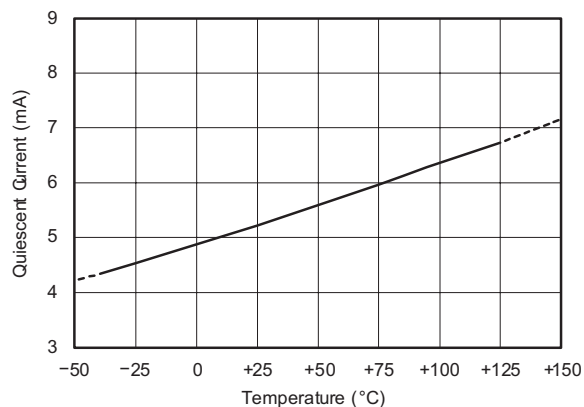
TYPICAL CHARACTERISTICS: $V_S = 2.8\text{ V}$ (continued)

At $T_A = +25^\circ\text{C}$ and $R_L = 150$, unless otherwise noted.

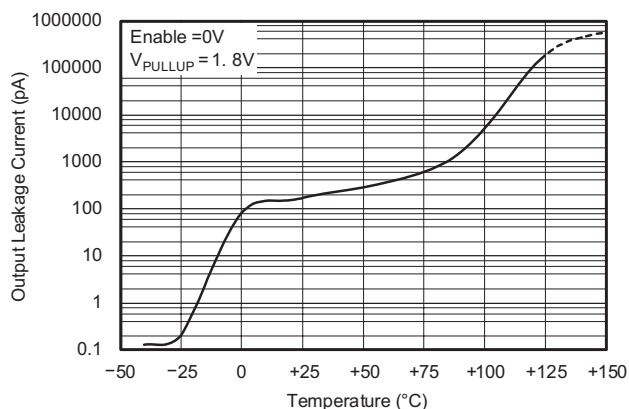
**V_{OUT} LEVEL SHIFT
vs
TEMPERATURE**



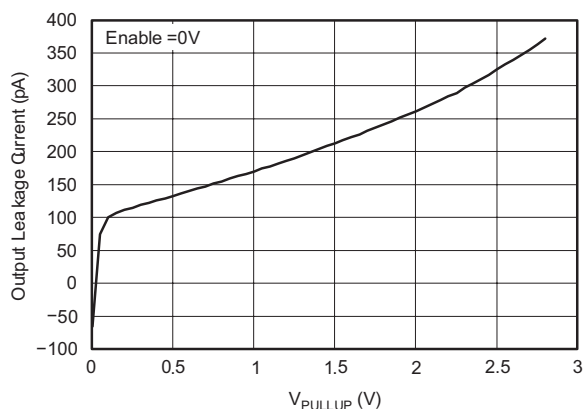
**QUIESCENT CURRENT
vs
TEMPERATURE**



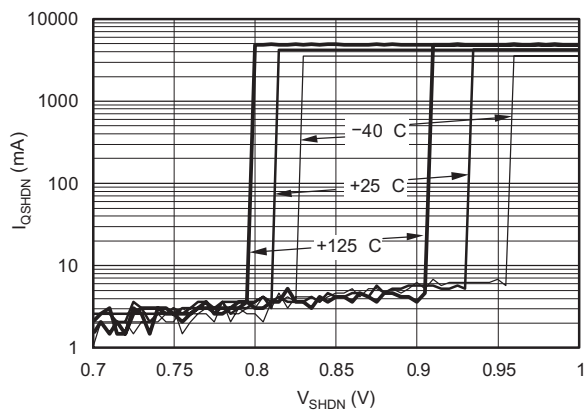
**OUTPUT LEAKAGE CURRENT
vs
TEMPERATURE**



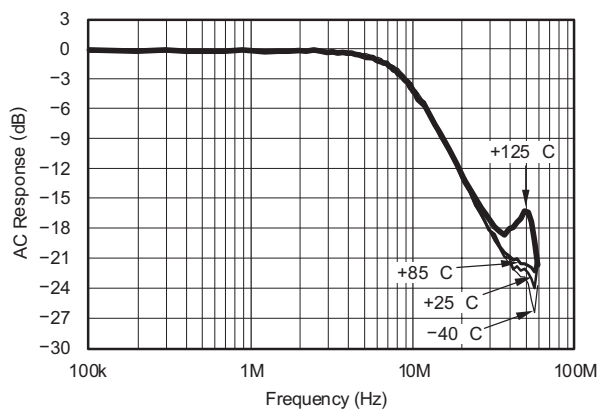
**OUTPUT LEAKAGE
vs
IN SHUTDOWN**



**SHUTDOWN QUIESCENT CURRENT HYSTERESIS
vs
TEMPERATURE**



**AC RESPONSE
vs
AT VARIOUS TEMPERATURES**



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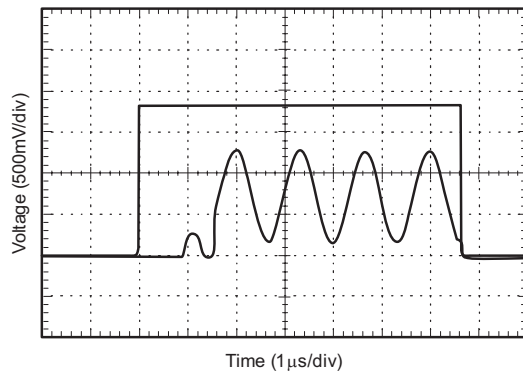
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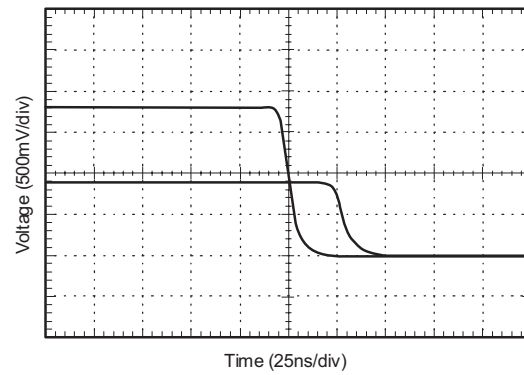
TYPICAL CHARACTERISTICS: $V_S = 2.8\text{ V}$ (continued)

At $T_A = +25^\circ\text{C}$ and $R_L = 150$, unless otherwise noted.

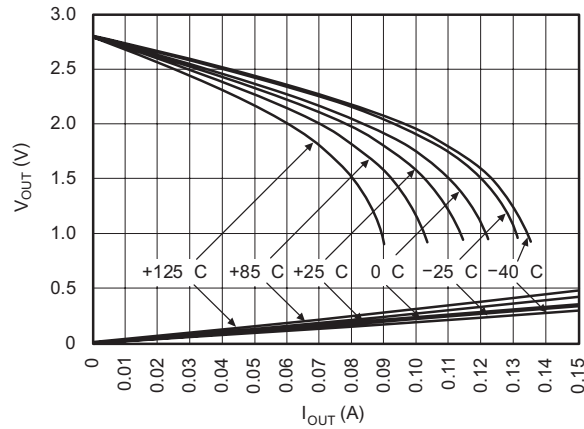
TURN-ON TIME



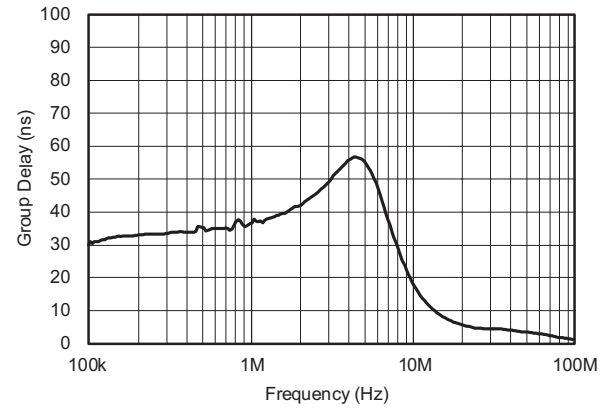
TURN-OFF TIME



**OUTPUT VOLTAGE
vs
OUTPUT CURRENT**



**GROUP DELAY
vs
FREQUENCY**



DIFFERENTIAL GAIN

INP = A - C SYNC = INT MTIME = 10s LINE = 330		
DG1	-0.6	%1
DG2	-1.0	%
DG3	-1.1	%
DG4	-1.2	%
DG5	-0.8	%5
STEPS 4 5 ZOOM 0 1 2 MODE 1		

DIFFERENTIAL PHASE

IN P = A - C SYNC = INT MTIME = 10s LINE = 330		
DP1	1.1	dg1
DP2	1.6	dg.
DP3	1.6	dg.
DP4	1.5	dg.
DP5	1.1	dg5
STEPS 4 5 ZOOM 0 1 2 MODE 1		

APPLICATION INFORMATION

The OPA361-Q1 video amplifier has been optimized to fit seamlessly with Texas Instruments' OMAP242x Multimedia processor. The following features have been integrated to provide excellent video performance.

- Internal gain setting resistors ($G = 5.2V/V$) reduce the number of external components needed in the video circuit.
- Integration of the 500Ω video encoder load resistor and $4k\Omega$ RSET resistor used by the OMAP242x helps minimize the number of external components and also ensures excellent ratio and temperature tracking. This feature helps to keep the overall gain accurate and stable over temperature.
- TV-detection support in connection with an OMAP242x multimedia processor. This feature helps to automate start/stop operation of the TV-out function and minimizes power consumption.
- A 2-pole filter is incorporated for DAC signal reconstruction.
- The OPA361-Q1 employs an internal level shift circuit that avoids sync pulse clipping and allows DC-coupled output.
- A shutdown feature reduces quiescent current to less than $1.5\mu A$ —crucial for portable applications

Although OPA361-Q1 is optimized for the OMAP242x processor, it is also suitable to interface with any digital media processor that outputs a video signal on the order of $0.4V_{PP}$ to $0.5V_{PP}$.

Figure 1 shows a typical application drawing with the OMAP242x processor and the TWL92230 Energy Management Chip.

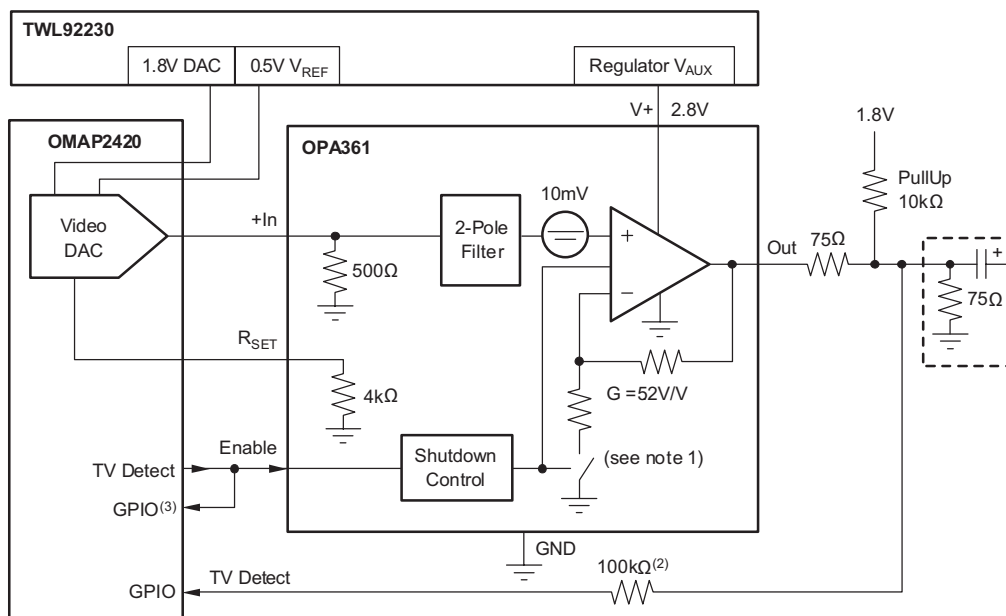


Figure 1. Typical Application using the OMAP242x and the TWL92230

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OPERATING VOLTAGE

The OPA361-Q1 is fully specified from 2.5V to 3.3V over a temperature range of -40°C to $+125^{\circ}\text{C}$. Parameters that vary significantly with operating voltages or temperature are shown in the Typical Characteristics. Power-supply pins should be bypassed with 100nF ceramic capacitors.

INPUT VOLTAGE

The input common-mode range of the OPA361-Q1 series extends from GND to 0.55V on a 3.3V supply. The input range is limited by the internal gain in conjunction with the maximum output swing capability and the power-supply voltage.

INPUT OVERVOLTAGE PROTECTION

All OPA361-Q1 pins are static-protected with internal ESD protection diodes connected to the supplies. These diodes will provide input overdrive protection if the current is externally limited to 10mA.

ENABLE/SHUTDOWN

The OPA361-Q1 has a shutdown feature that disables the output and reduces the quiescent current to less than 1.5 μA . This feature is especially useful for portable video applications, where the device is infrequently connected to a television (TV) or other video device.

The Enable logic input voltage is referenced to the OPA361-Q1 GND pin. A logic level HIGH applied to the enable pin enables the op amp. The logic levels are compatible with 1.8V CMOS logic levels. A valid logic HIGH is defined as $> 1.3\text{V}$ above GND. A valid logic LOW is defined as $< 0.35\text{V}$ above GND. If the Enable pin is not connected, internal pull-up circuitry will enable the amplifier.

When disabling the OPA361-Q1, internal circuitry also disconnects the internal gain setting feedback. This feature is in support of the TV-detect function. See the *TV-Detect Function* section for more detailed information.

INTERNAL 2-POLE FILTER

The OPA361-Q1 filter is a Sallen-Key topology with a 9MHz cutoff frequency. Figure 2 shows a detailed drawing of the filter components. This filter allows video signals to pass without any visible distortion, as shown in Figure 3 through Figure 6. The video encoder embedded in the OMAP242x processor typically samples at 54MHz. At this frequency, the attenuation is typically 23dB, which effectively attenuates the sampling aliases.

The internal 500 Ω resistor on the input to GND converts the output current of the OMAP2420 internal video DAC into a voltage. It is also part of the Sallen-Key filter. Using an external resistor to adjust the input voltage range will also alter the filter characteristics.

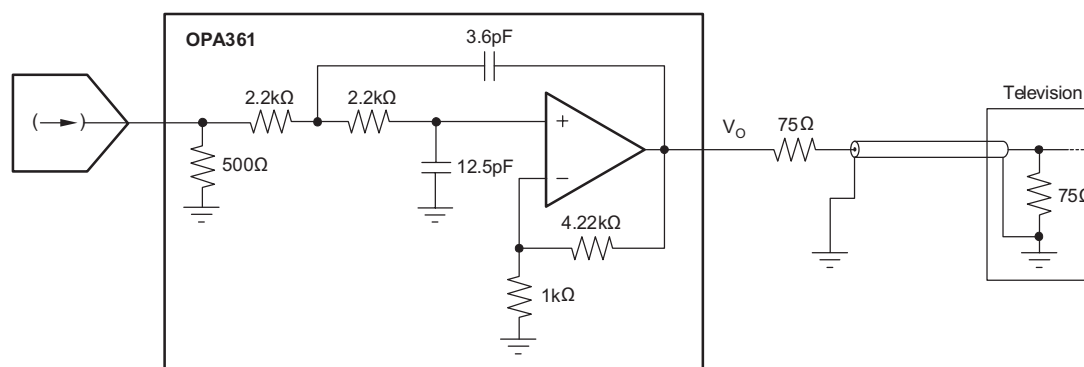


Figure 2. Filter Structure of the OPA361-Q1

Video Performance

The color bar signal in [Figure 3](#) shows excellent amplitude characteristics and no attenuation of colors with respect to the luminance signal.

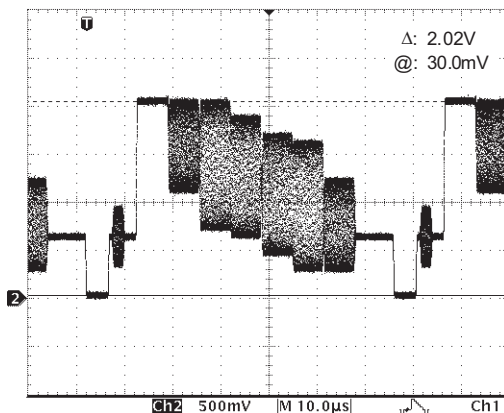


Figure 3. 100/75 Color Bar Signal at Output of OPA361-Q1

The CCIR330/5 test pattern requires one of the greatest dynamic ranges, and therefore tests the OPA361-Q1 output voltage swing capability. The scope plot shown in [Figure 4](#) has been taken with a 2.8V supply and shows no clipping on the top side of the signal.

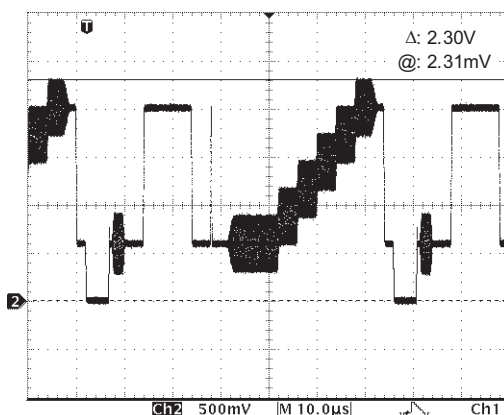


Figure 4. CCIR330/5: No Clipping, Even On 2.8V Supply

The multiburst test patterns have different sine-wave burst sections with the following frequencies: 0.5MHz, 1MHz, 2MHz, 4MHz, 4.8MHz and 5.8MHz with 420mVPP. There is no visible attenuation even at the highest frequencies, which indicates a very flat frequency response of the OPA361-Q1. As shown in [Figure 5](#) and [Figure 6](#), the top line illustrates the full signal and the bottom line is a more detailed view of the last three sine wave bursts.

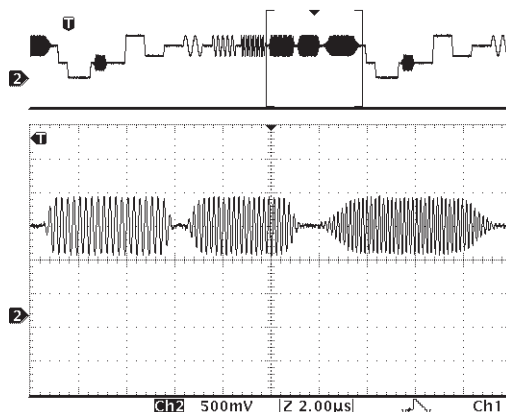


Figure 5. Multiburst Signal (CCIR 18/1) Shows Very Flat Frequency Response

The CCIR17 test pattern contains a 2T and a 20T pulse, as shown in [Figure 6](#). The 2T pulse is used to check for pulse distortion and reflection, and the 20T pulse is used to check for amplitude and group delay between chrominance and luminance. Neither pulse exhibits any distortion or group delay artifacts.

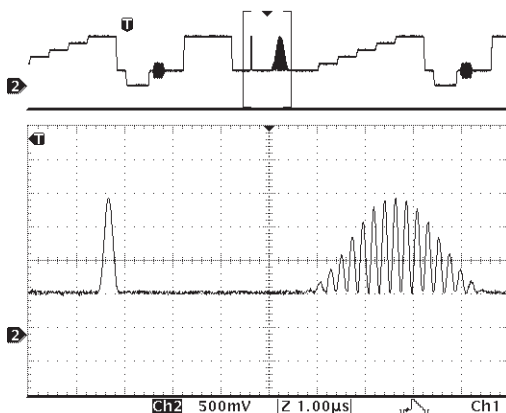


Figure 6. CCIR 17 2T and 20T Pulses Show No Visible Distortion

INTERNAL LEVEL SHIFT

Many common video DACs embedded in digital media processors, like the new OMAP242x processors, operate on a single supply (no negative supply). Typically, the lowest point of the sync pulse output by these video DACs is close to 0V. With a 0V input, the output of a common single-supply op amp saturates at a voltage > 0V. This effect would clip the sync pulse, and therefore degrade the video signal integrity. The OPA361-Q1 employs an internal level shift circuit to avoid clipping. The input signal is typically shifted by approximately 11mV. This shift is well within the linear output voltage range of the OPA361-Q1 with a standard 150Ω video load.

Output Swing Capability

[Figure 7](#) shows the true output swing capability of the OPA361-Q1 by taking the tip of the input sync pulse to a slightly negative voltage. Even when the output sync tip is at 3mV, the output after the 75Ω series termination still shows no clipping of the sync pulse.

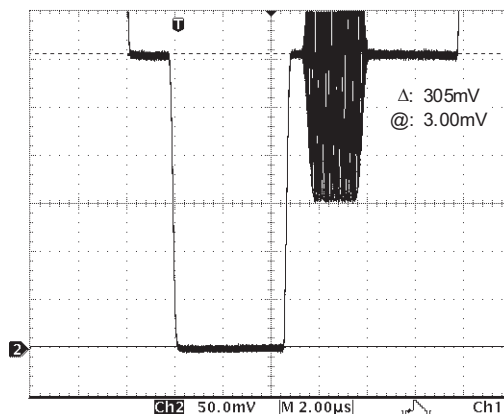


Figure 7. No Clipping of the Sync Pulse

TV-Detect Function

The TV-detection feature of the OPA361-Q1 works in conjunction with the OMAP242x (or other processors) to detect if a television is connected to the video output of the device. In order to detect a TV load, the OPA361-Q1 is briefly turned off, ideally during the first vertical sync pulse. For the detection, a simple pull-up resistor to the processor logic supply is used on the output of the OPA361-Q1. The voltage level is pulled LOW if the TV (or other video equipment) is connected, or HIGH if nothing is connected. A GPIO in the processor can be used to read this logic level and decide if a video load is connected. Figure 8 shows a scope plot with the TV disconnected and Figure 9 shows a scope plot with the TV connected; the upper line in both figures is the disable pulse. Figure 10 shows a circuit drawing using the TV-detect signal to disable or enable the OPA361-Q1.

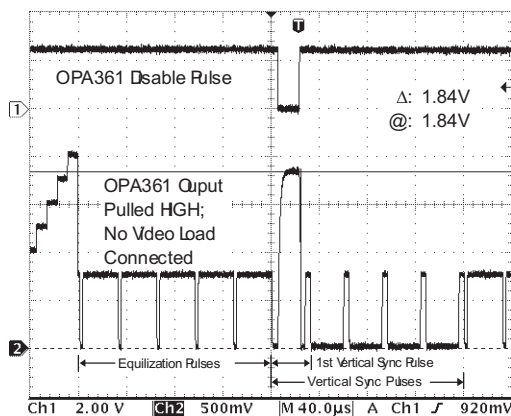


Figure 8. Output of OPA361-Q1 Pulled Up To 1.8V During Disable: TV Disconnected

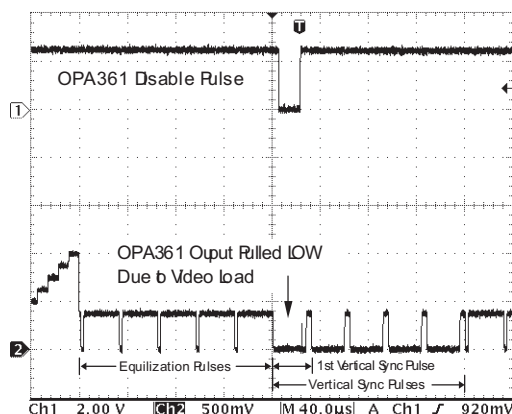


Figure 9. Output of OPA361-Q1 Pulled Down: TV Connected.

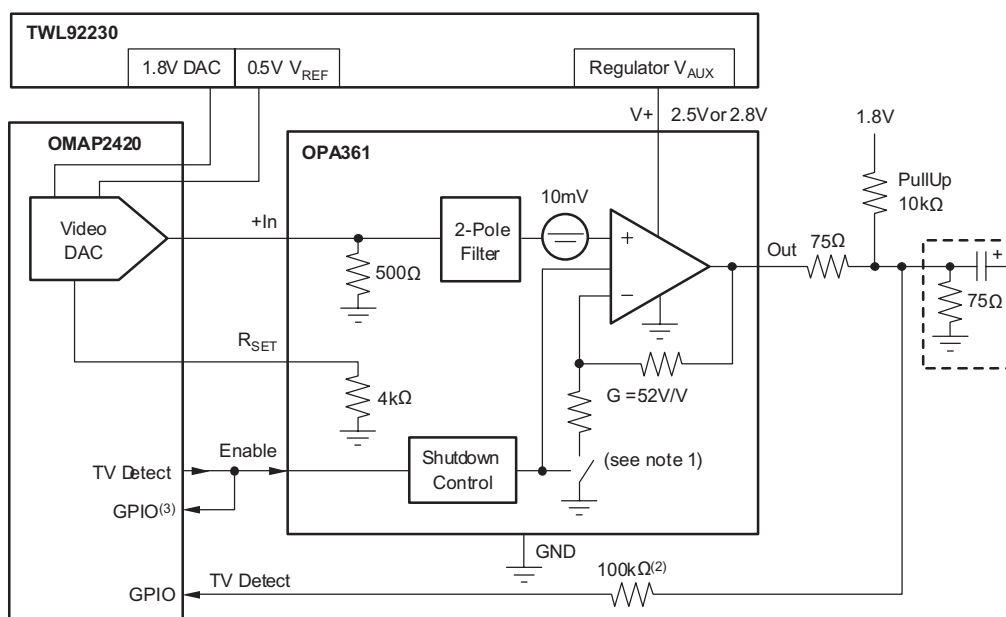


Figure 10. Using TV-Detect Signal to Disable/Enable the OPA361-Q1

Disabling the OPA361-Q1 also disconnects the internal feedback resistors' path to GND, and therefore there is no current flowing from the logic supply through the pull-up resistor to GND if no video load is connected; this helps to conserve battery life. The typical leakage when the output is pulled high and OPA361-Q1 is disabled is only about 300pA.

The following functionality can be achieved by implementing TV-detection:

- Automatic video start by polling the video line periodically.
- Automatic video stop if the TV (or other equipment) is disconnected.

Proper implementation allows to significantly simplify the user interface.

For more information, see Application Report SBOA109, *OPA361-Q1 and TV Detection*, available for download at www.ti.com.

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA361AQDCKRQ1	ACTIVE	SC70	DCK	6	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	QXP	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

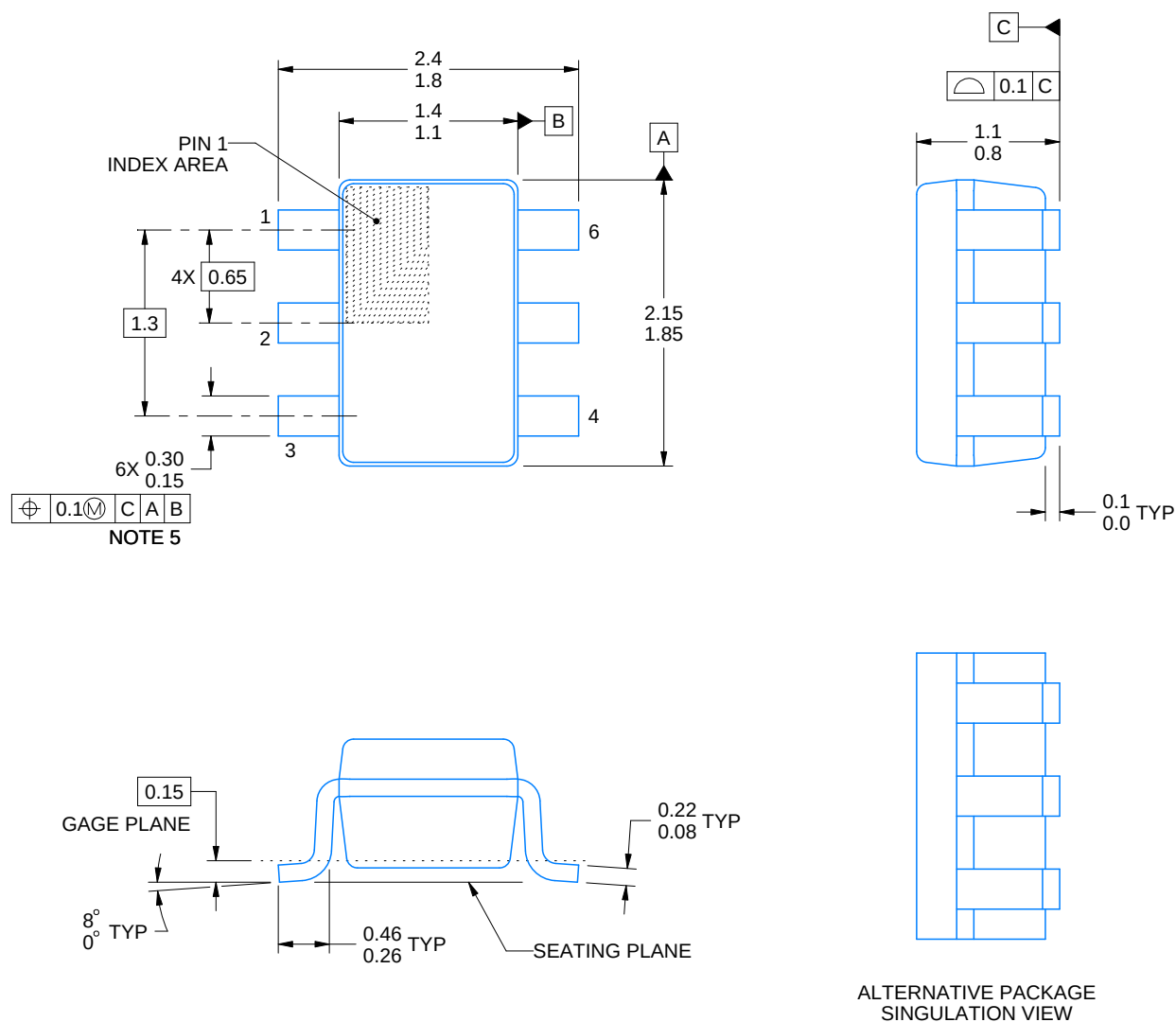
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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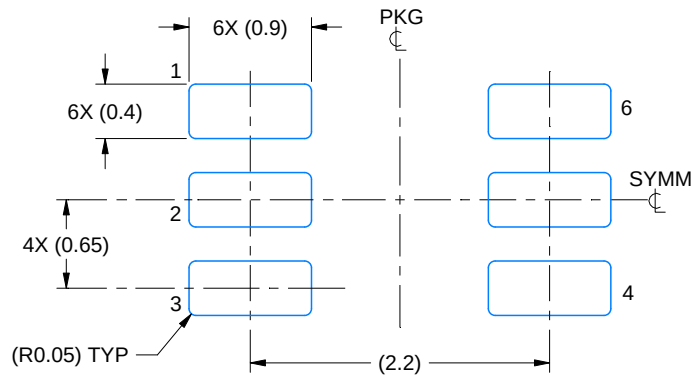
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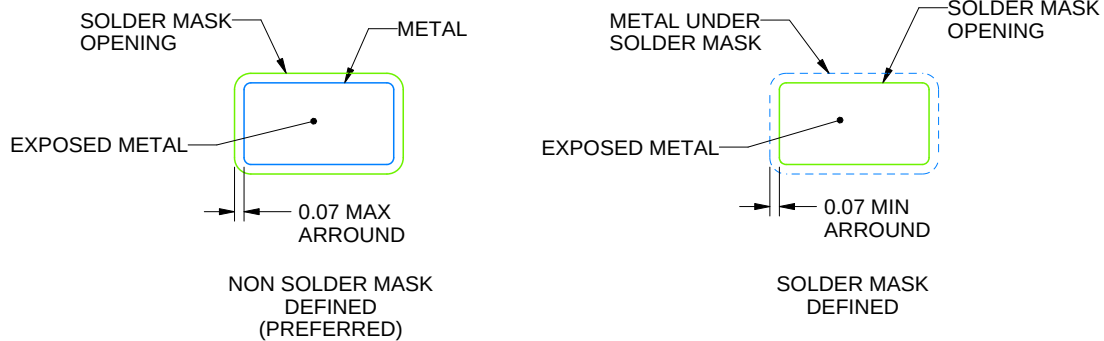
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NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
4. Falls within JEDEC MO-203 variation AB.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:18X

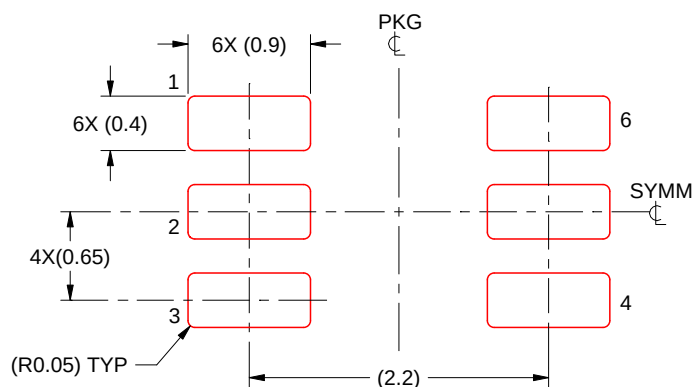


SOLDER MASK DETAILS

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NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.
6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SOLDER PASTE EXAMPLE
 BASED ON 0.125 THICK STENCIL
 SCALE:18X

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NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

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